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# FAN48610 2.5 MHz, Fixed-Output Synchronous TinyBoost® Regulator

### Features

- Input Voltage Range: 2.5 V to 4.50 V
- Output Voltages Range: 3.0 V to 5.0 V
- I<sub>OUT</sub> ≥ 1 A at V<sub>OUT</sub> = 5.0 V, V<sub>IN</sub> ≥ 2.5 V
- I<sub>OUT</sub>  $\geq$  1.5 A at V<sub>OUT</sub> = 5.0 V, V<sub>IN</sub>  $\geq$  3.0 V
- Up to 94% Efficient
- Internal Synchronous Rectification
- Soft-Start with True Load Disconnect
- Short-Circuit Protection
- 9-Bump, 1.215 mm x 1.215 mm, 0.4 mm Pitch WLCSP
- Three External Components: 2016 0.47 µH Inductor, 0603 Case Size Input / Output Capacitors
- Total Application Board Solution Size: < 11 mm<sup>2</sup>

## Applications

- Class-D Audio Amplifier and USB OTG Supply
- Boost for Low-Voltage Li-Ion Batteries
- Smart Phones, Tablets, Portable Devices, Wearables

### Description

The FAN48610 is a low-power boost regulator designed to provide a minimum voltage-regulated rail from a standard single-cell Li-lon battery and advanced battery chemistries. Even below the minimum system battery voltage, the device maintains the output voltage regulation for a minimum output load current of 1.0 A. The combination of built-in power transistors, synchronous rectification, and low supply current suit the FAN48610 for battery-powered applications.

The FAN48610 is available in a 9-bump, 0.4 mm pitch, Wafer-Level Chip-Scale Package (WLCSP).

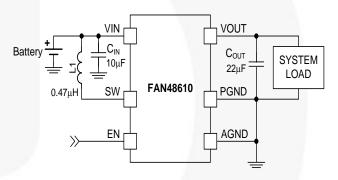


Figure 1. Typical Application

### Ordering Information

-					
Part Number	V <sub>OUT</sub>	Operating Temperature	Package	Packing <sup>(1)</sup>	Device Marking
FAN48610UC50X					KE
FAN48610BUC50X <sup>(2)</sup>	5.0 V		-40°C to 85°C WLCSP, 0.4 mm Pitch Tape and Reel	Topo and Dool	KF
FAN48610BUC45X <sup>(2)</sup>	4.5 V	-40°C to 85°C		Tape and Reel	KA
FAN48610BUC33X <sup>(2)</sup>	3.3 V				KN

#### Notes:

1. Tape and reel specifications are available on www.fairchildsemi.com.

2. Includes backside lamination.

### **Block Diagrams**

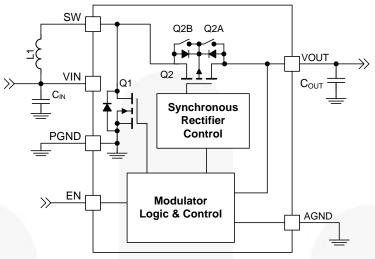


Figure 2. IC Block Diagram

### Table 1. Recommended Components

Component	Description	Vendor	Parameter	Тур.	Unit
		Toko: DFE201612C	L	0.47	μH
L1	0.47 µH, 30%, 2016	DFR201612C Cyntec: PIFE20161B	DCR (Series R)	40	mΩ
C <sub>IN</sub>	10 µF, 10%, 6.3 V, X5R, 0603	Murata: GRM188R60J106K TDK: C1608X5R0J106K	С	10	μF
C <sub>OUT</sub>	22 µF, 20%, 6.3 V, X5R, 0603	TDK: C1608X5R0J226M	С	22	μF

# **Pin Configuration**

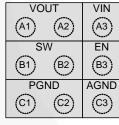
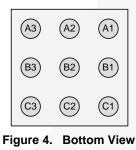


Figure 3. Top View



### **Pin Definitions**

Pin #	Name	Description
A1, A2	VOUT	Output Voltage. This pin is the output voltage terminal; connect directly to COUT.
A3	VIN	Input Voltage. Connect to Li-lon battery input power source and the bias supply for the gate drivers.
B1, B2	SW	Switching Node. Connect to inductor.
B3	EN	Enable. When this pin is HIGH, the circuit is enabled.
C1, C2	PGND	<b>Power Ground</b> . This is the power return for the IC. C <sub>OUT</sub> capacitor should be returned with the shortest path possible to these pins.
C3	AGND	<b>Analog Ground</b> . This is the signal ground reference for the IC. All voltage levels are measured with respect to this pin – connect to PGND at a single point.

### **Absolute Maximum Ratings**

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

Symbol	Parameter			Max.	Unit
V <sub>IN</sub>	Voltage on VIN Pin		-0.3	6.0	V
Vout	Voltage on VOUT Pin			6.0	V
CW/	SW/ Node	DC	-0.3	6.0	V
SW	SW Node	Transient: 10 ns, 3 MHz	-1.0	8.0	V
V <sub>CC</sub>	Voltage on Other Pins			6.0 <sup>(3)</sup>	V
500	Human Body Model per JESD22-A114		4 2		kV
ESD	Electrostatic Discharge Protection Level Charged Device Model per JESD22-C101			1	
TJ	Junction Temperature		-40	+150	°C
T <sub>STG</sub>	Storage Temperature		-65	+150	°C
T∟	Lead Soldering Temperature, 10 Seconds	Lead Soldering Temperature, 10 Seconds			°C

Note:

3. Lesser of 6.0 V or  $V_{IN}$  + 0.3 V.

### **Recommended Operating Conditions**

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to absolute maximum ratings.

Symbol	Parameter	Min.	Max.	Unit
V <sub>IN</sub>	Supply Voltage	2.5	4.5	V
I <sub>OUT</sub>	Maximum Output Current	1000		mA
T <sub>A</sub>	Ambient Temperature	-40	+85	°C
TJ	Junction Temperature	-40	+125	°C

### **Thermal Properties**

Junction-to-ambient thermal resistance is a function of application and board layout. This data is measured with four-layer 2s2p boards with vias in accordance to JEDEC standard JESD51. Special attention must be paid not to exceed junction temperature,  $T_{J(max)}$ , at a given ambient temperature,  $T_A$ .

Symbol	Parameter	Typical	Unit
Θ <sub>JA</sub>	Junction-to-Ambient Thermal Resistance	50	°C/W

## **Electrical Characteristics**

Recommended operating conditions, unless otherwise noted, circuit per Figure 1,  $V_{OUT}$ = 3.0 V to 5.0 V,  $V_{IN}$  = 2.5 V to 4.5 V,  $T_A$  = -40°C to 85°C. Typical values are given  $V_{IN}$  = 3.6 V and  $T_A$  = 25°C.

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit
Power Sup	oply			1		<u> </u>
		V <sub>IN</sub> =3.6 V, I <sub>OUT</sub> =0, EN= V <sub>IN</sub>		85	125	
Ι <sub>Q</sub>	V <sub>IN</sub> Quiescent Current	Shutdown: EN=0, V <sub>IN</sub> =3.6 V		3	10	μA
V <sub>UVLO</sub>	Under-Voltage Lockout	V <sub>IN</sub> Rising		2.2	2.3	V
V <sub>UVLO_HYS</sub>	Under-Voltage Lockout Hysteresis			150		mV
Inputs			•	•		
VIH	Enable HIGH Voltage		1.05			V
V <sub>IL</sub>	Enable LOW Voltage				0.4	V
I <sub>PD</sub>	Current Sink Pull-Down	EN Pin, Logic HIGH		100		nA
R <sub>LOW</sub>	Low-State Active Pull-Down	EN Pin, Logic LOW	200	300	400	kΩ
Outputs			•			
V <sub>REG</sub>	Output Voltage Accuracy DC <sup>(4)</sup>	Referred to $V_{OUT}$ , 2.5 V $\leq V_{IN} \leq V_{OUT}$ -150 mV	-2		4	%
I <sub>LK_OUT</sub>	VIN -to-VOUT Leakage Current	V <sub>OUT</sub> =0, EN=0, V <sub>IN</sub> =4.2 V			1	μA
I <sub>LK</sub>	VOUT-to-VIN Reverse Leakage Current	V <sub>OUT</sub> =5.0 V, EN=0, V <sub>IN</sub> =2.5 V			3.5	μA
VTRSP	Output Voltage Accuracy Transient <sup>(5)</sup>	Referred to VOUT, 50-500 mA Load Step	-5		5	%
Timing				•		
f <sub>SW</sub>	Switching Frequency	V <sub>IN</sub> =3.6 V, V <sub>OUT</sub> =5.0 V, Load=1000 mA	2.0	2.5	3.0	MHz
t <sub>SS</sub>	Soft-Start EN HIGH to Regulation <sup>(5)</sup>	50 Ω Load, V <sub>OUT</sub> = 5.0 V		600		μS
t <sub>RST</sub>	FAULT Restart Timer <sup>(5)</sup>			20		ms
Power S	tage					
R <sub>DS(ON)N</sub>	N-Channel Boost Switch R <sub>DS(ON)</sub>	V <sub>IN</sub> =3.6 V, V <sub>OUT</sub> =5.0 V		80	130	mΩ
R <sub>DS(ON)P</sub>	P-Channel Sync. Rectifier R <sub>DS(ON)</sub>	V <sub>IN</sub> =3.6 V, V <sub>OUT</sub> =5.0 V		65	115	mΩ
Iv_lim	Boost Valley Current Limit	V <sub>OUT</sub> =5.0 V		3.0		Α
I <sub>V_LIM_SS</sub>	Boost Soft-Start Valley Current Limit	V <sub>IN</sub> <v<sub>OUT &lt; V<sub>OUT_TARGET</sub>, SS Mode</v<sub>	1	1.7		А
V <sub>MIN_1.0A</sub>	Minimum V <sub>IN</sub> for 1000 mA Load <sup>(5)</sup>	V <sub>OUT</sub> =5.0 V		2.5		V
V <sub>MIN_1.5A</sub>	Minimum V <sub>IN</sub> for 1500 mA Load <sup>(5)</sup>	V <sub>OUT</sub> =5.0 V		3.0		V
T <sub>150T</sub>	Over-Temperature Protection (OTP)			150		°C
T <sub>150H</sub>	OTP Hysteresis			20		°C

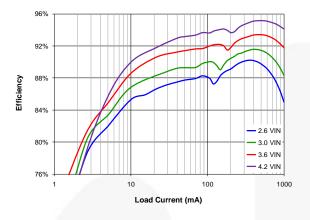
#### Notes:

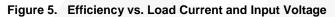
4. DC I<sub>LOAD</sub> from 0 to 1 A. V<sub>OUT</sub> measured from mid-point of output voltage ripple. Effective capacitance of  $C_{OUT} \ge 3 \mu F$ .

5. Guaranteed by design and characterization; not tested in production.

## **Typical Characteristics**

Unless otherwise specified;  $V_{IN} = 3.6 V$ ,  $V_{OUT} = 5.0 V$ ,  $T_A = 25^{\circ}C$ , and circuit and components according to Figure 1.





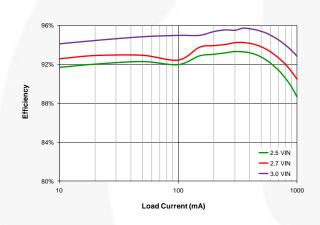


Figure 7. Efficiency vs. Load Current and Input Voltage, Vout=3.3V

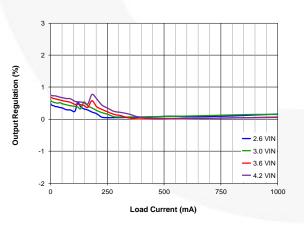


Figure 9. Output Regulation vs. Load Current and Input Voltage (Normalized to 3.6 VIN, 500 mA Load)

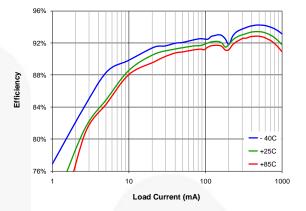


Figure 6. Efficiency vs. Load Current and Temperature

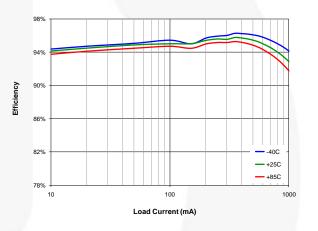
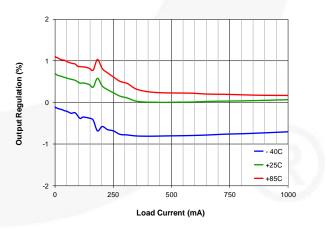
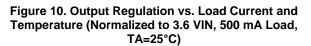


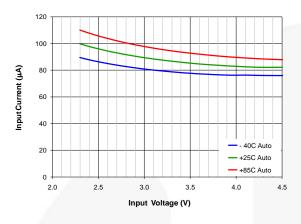
Figure 8. Efficiency vs. Load Current and Temperature,  $V_{\text{IN}}{=}3.0V,\,V_{\text{OUT}}{=}3.3V$ 



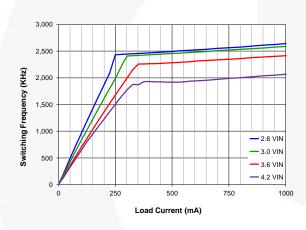


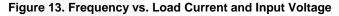
### **Typical Characteristics**

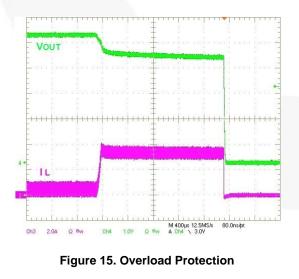
Unless otherwise specified;  $V_{IN} = 3.6 \text{ V}$ ,  $V_{OUT} = 5.0 \text{ V}$ ,  $T_A = 25^{\circ}C$ , and circuit and components according to Figure 1.











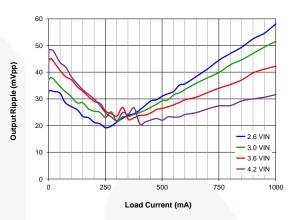
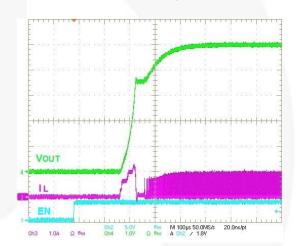
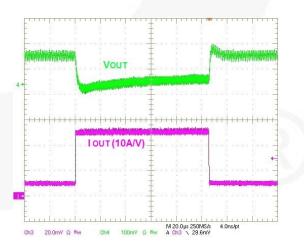
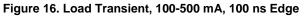


Figure 12. Output Ripple vs. Load Current and Input Voltage









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## **Typical Characteristics** Unless otherwise specified; V<sub>IN</sub> = 3.6 V, V<sub>QUT</sub> = 5.0 V, T<sub>A</sub> = 25°C, and circuit and components according to Figure 1. VOUT VIN I OUT (10A/V) VOUT M 20.0µs 250MS/s A Ch3 \ 75.0mV 4.0ns/pt 50.0mV Ω Bw ŶΩ Ch3 50.0mV Ω By 200mV Ω Figure 17. Load Transient, 500-1000 mA, 100 ns Edge

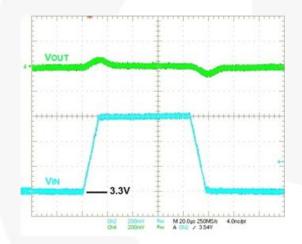


Figure 19. Line Transient, 3.3-3.9 V<sub>IN</sub>, 10 μs Edge, 500 mA Load

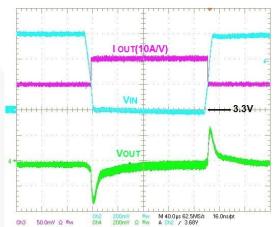


Figure 18. Simultaneous Line / Load Transient, 3.3-3.9 VIN, 10 µs Edge, 500-1000 mA Load, 100 ns Edge

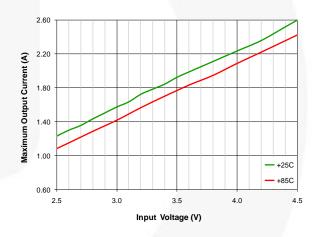


Figure 20. Typical Maximum Output Current vs. Input Voltage

### **Circuit Description**

FAN48610 is a synchronous boost regulator, typically operating at 2.5 MHz in Continuous Conduction Mode (CCM), which occurs at moderate to heavy load current and low V<sub>IN</sub> voltages. The regulator's Pass-Through Mode automatically activates when V<sub>IN</sub> is above the boost regulator's set point.

Mode	Description	Invoked When:				
LIN	Linear Startup	$V_{IN} > V_{OUT}$				
SS	Boost Soft-Start	$V_{IN} < V_{OUT} < V_{OUT(TARGET)}$				
BST	Boost Operating Mode	$V_{OUT} = V_{OUT(TARGET)}$				
PT	Pass-Through Mode	$V_{IN} > V_{OUT(TARGET)}$				

Table	2.	Operating	Modes
Iabic	<u> </u>	operating	mouca

### **Boost Mode Regulation**

The FAN48610 uses a current-mode modulator to achieve excellent transient response and smooth transitions between CCM and DCM operation. During CCM operation, the device maintains a switching frequency of about 2.5 MHz. In light-load operation (DCM), frequency is naturally reduced to maintain high efficiency.

#### Shutdown and Startup

When EN is LOW, all bias circuits are off and the regulator is in Shutdown Mode. During shutdown, current flow is prevented from VIN to VOUT, as well as reverse flow from VOUT to VIN. It is recommended to keep load current draw below 500 mA until the devices successfully executes startup. The following table describes the startup sequence.

Start Mode	Entry	Exit	End Mode	Timeout (µs)
LIN1	V <sub>IN</sub> > V <sub>UVLO</sub> ,	V <sub>OUT</sub> > V <sub>IN</sub> - 300 mV	SS	
	EN=1	TIMEOUT	LIN2	512
LIN2	LIN1 Exit	V <sub>OUT</sub> > V <sub>IN</sub> - 300 mV	SS	
		TIMEOUT	FAULT	1024
	LIN1 or	Vout=Vout(target)	BST	
SS	LIN1 Of LIN2 Exit	OVERLOAD TIMEOUT	FAULT	64

#### Table 3. Boost Startup Sequence

### LIN Mode

When EN is HIGH and  $V_{IN} > V_{UVLO}$ , the regulator first attempts to bring  $V_{OUT}$  within 300 mV of  $V_{IN}$  by using the internal fixed-current source from VIN (Q2). The current is limited to the LIN1 set point.

If  $V_{\text{OUT}}$  reaches  $V_{\text{IN}}\text{-}300$  mV during LIN1 Mode, the SS Mode is initiated. Otherwise, LIN1 times out after 512  $\mu s$  and LIN2 Mode is entered.

In LIN2 Mode, the current source is incremented to 1.6 A. If  $V_{\text{OUT}}$  fails to reach  $V_{\text{IN}}\text{-}300\,\text{mV}$  after 1024  $\mu\text{s},$  a fault condition is declared and the device waits 20 ms to attempt an automatic restart.

#### Soft-Start (SS) Mode

Upon the successful completion of LIN Mode ( $V_{OUT} \ge V_{IN}$ -300 mV), the regulator begins switching with boost pulses current limited to 50% of nominal level.

During SS Mode, if  $V_{OUT}$  fails to reach regulation during the SS ramp sequence for more than 64  $\mu$ s, a fault is declared. If large  $C_{OUT}$  is used, the reference is automatically stepped slower to avoid excessive input current draw.

#### Boost (BST) Mode

This is a normal operating mode of the regulator.

#### Pass-Through (PT) Mode

In normal operation, the device automatically transitions from Boost Mode to Pass-Through Mode if V<sub>IN</sub> goes above the target V<sub>OUT</sub>. In Pass-Through Mode, the device fully enhances Q2 to provide a very low impedance path from VIN to VOUT. Entry to the Pass-Through Mode is triggered by condition where V<sub>IN</sub> > V<sub>OUT</sub> and no switching has occurred during the past 5  $\mu$ s. To soften the entry into Pass-Through Mode, Q2 is driven as a linear current source for the first 5  $\mu$ s. Pass-Through Mode exit is triggered when V<sub>OUT</sub> reaches the target V<sub>OUT</sub> voltage. During Automatic Pass-Through Mode, the device is short-circuit protected by a voltage comparator tracking the voltage drop from V<sub>IN</sub> to V<sub>OUT</sub>; if the drop exceeds 300 mV, a fault is declared.

### **Fault State**

The regulator enters Fault State under any of the following conditions:

- V<sub>OUT</sub> fails to achieve the voltage required to advance from LIN Mode to SS Mode.
- V<sub>OUT</sub> fails to achieve the voltage required to advance from SS Mode to BST Mode.
- Boost current limit triggers for 2 ms during BST Mode.
- V<sub>IN</sub> V<sub>OUT</sub> > 300 mV; this fault can occur only after successful completion of the soft-start sequence.
- $V_{IN} < V_{UVLO}$ .

Once a fault is triggered, the regulator stops switching and presents a high-impedance path between VIN and VOUT. After waiting 20 ms, an automatic restart is attempted.

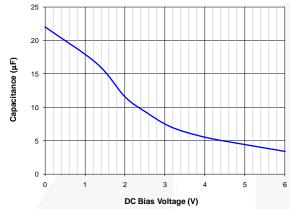
#### **Over-Temperature**

The regulator shuts down if the die temperature exceeds 150°C. Restart occurs when the IC has cooled by approximately 20°C.

### **Application Information**

### **Output Capacitance (COUT)**

The effective capacitance  $(C_{EFF}^{(6)})$  of small, high-value ceramic capacitors decreases as their bias voltage increases, as illustrated in the graph below:



#### Figure 21. C<sub>EFF</sub> for 22 µF, 0603, X5R, 6.3 V-Rated Capacitor (TDK C1608X5R0J226M)

FAN48610 is guaranteed for stable operation with the minimum value of  $C_{EFF}$  ( $C_{EFF(MIN)}$ ) outlined in Table 4.

Table 4. Minimum C<sub>EFF</sub> Required for Stability

Оре	Operating Conditions				
<b>V</b> <sub>оυт</sub> (V)	V <sub>OUT</sub> (V) V <sub>IN</sub> (V) I <sub>LOAD</sub> (mA)				
5.0	2.5 to 4.5	0 to 1000	3.0		

Note:

C<sub>EFF</sub> varies by manufacturer, capacitor material, and 6. case size.

### Inductor Selection

Recommended nominal inductance value is 0.47 µH.

The FAN48610 employs valley-current limiting, so peak inductor current can reach 3.8 A for a short duration during overload conditions. Saturation effects cause the inductor current ripple to become higher under high loading, as only the valley of the inductor current ripple is controlled.

### Startup

Input current limiting is in effect during soft-start, which limits the current available to charge  $C_{OUT}$  and any additional capacitance on the V<sub>OUT</sub> line. If the output fails to achieve regulation within the limits described in the Soft-Start section above, a fault occurs, causing the circuit to shut down. It waits about 20 ms before attempting a restart. If the total combined output capacitance is very high, the circuit may not start on the first attempt, but eventually achieves regulation if no load is present. If a high current load and high capacitance are both present during soft-start, the circuit may fail to achieve regulation and continually attempt soft-start, only to have the output capacitance discharged by the load when in Fault State.

### **Output Voltage Ripple**

Output voltage ripple is inversely proportional to COUT. During ton, when the boost switch is on, all load current is supplied by COUT.

$$V_{RIPPLE(P-P)} = t_{ON} \bullet \frac{I_{LOAD}}{C_{OUT}}$$
EQ. 1

and

$$t_{ON} = t_{SW} \bullet D = t_{SW} \bullet \left(1 - \frac{V_{IN}}{V_{OUT}}\right)$$
 EQ. 2

therefore:

$$V_{RIPPLE(P-P)} = t_{SW} \bullet \left(1 - \frac{V_{IN}}{V_{OUT}}\right) \bullet \frac{I_{LOAD}}{C_{OUT}}$$
 EQ. 3

$$f_{SW} = \frac{1}{f_{SW}}$$
 EQ. 4

The maximum  $V_{RIPPLE}$  occurs when  $V_{IN}$  is minimum and  $I_{LOAD}$ is maximum. For better ripple performance, more output capacitance can be added.

### Layout Recommendations

The layout recommendations below highlight various topcopper pours by using different colors.

To minimize spikes at VOUT, COUT must be placed as close as possible to PGND and VOUT, as shown below.

For thermal reasons, it is suggested to maximize the pour area for all planes other than SW. Especially the ground pour should be set to fill all available PCB surface area and tied to internal layers with a cluster of thermal vias.

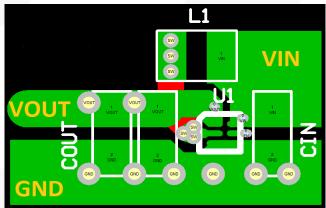


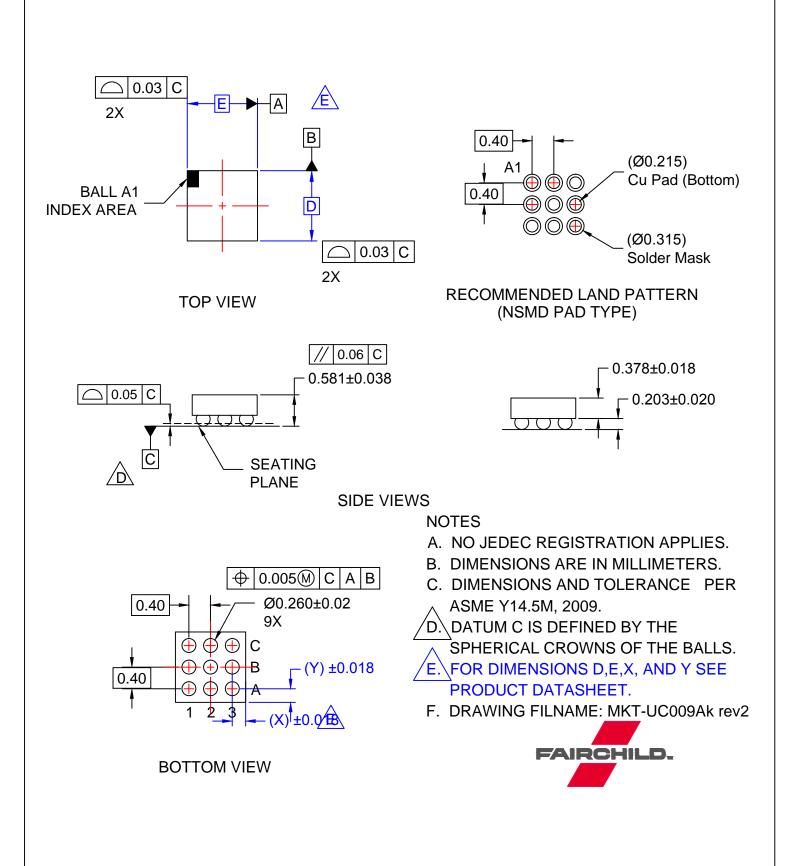
Figure 22. Layout Recommendation

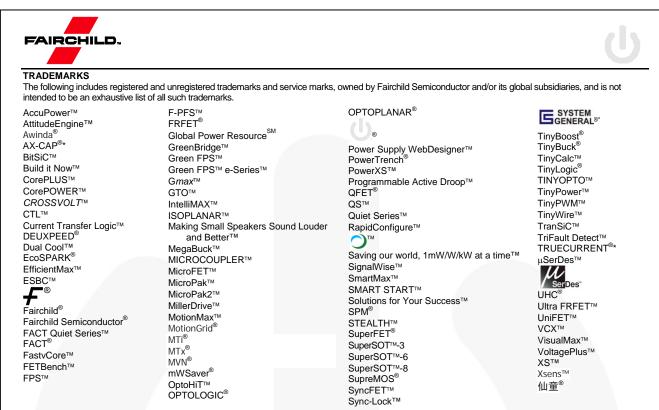
**Product-Specific Dimensions** (This table pertains to the package information on the following page.)

D	E	X	Y
1.215 ±0.030 mm	1.215 ±0.030 mm	0.2075 mm	0.2075 mm

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