

Adjustable Linear Low Dropout LED Driver

TLE4309



Features

- Adjustable constant current up to 500 mA
- Operating supply voltage range 4.5 V to 24 V
- Maximum supply voltage up to 45 V
- Low dropout voltage
- PWM / ENABLE input
- << 1 µA quiescent current when disabled
- Overtemperature protection
- Short circuit proof
- Reverse polarity protection
- Operating junction temperature range: -40 °C to 150 °C
- Suitable for applications up to 85 °C ambient temperature.
- Green Product (RoHS compliant)

Functional Description

The TLE4309 is an integrated adjustable constant current source for driving loads up to 500 mA. The output current level can be adjusted with an external shunt resistor. Supplying high power LEDs with the TLE4309 ensures constant brightness independent from supply voltage or LED forward voltage spread. Therefore, LED lifetime is extended by protecting from overcurrent and overtemperature

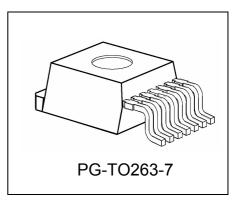
The PWM/EN input permits LED brightness regulation by pulse width modulation. Setting the pin to "low" switches off the IC entirely. Due to the high impedance of the PWM/EN input, the TLE4309 can be used as a protected high side switch.

Protection circuits prevent from damage to the IC in case of overload, short circuit, and reverse polarity. A chip temperature monitoring circuit shuts off the power stage and prevents the IC from destruction under fault conditions.

In case of negative input voltage, a leakage current is flowing only from the output to the input, hence the LED are protected against reverse supply. Input voltage peaks up to 45 V are absorbed by the IC, preventing the LEDs from overcurrent.

The TLE4309G is provided in the surface mounted PG-TO-263 package with excellent thermal resistance.

Туре	Package	Marking
TLE4309G	PG-TO263-7-1	TLE4309





Circuit Description

An external shunt resistor in the ground path of the connected LEDs senses the LED current. A regulation loop maintains the voltage drop across the shunt resistor at a constant level. Selecting the shunt resistance permits to adjust the appropriate constant current . The output current calculates

$$I_{\rm Q,\,typ} = \frac{V_{\rm REF}}{R_{\rm REF}} \tag{1}$$

where V_{REF} is the reference voltage. (see "Electrical Characteristics" table). The equation applies in a range of 0.39 $\Omega \le R_{\text{REF}} \le 1.8 \Omega$. On Page 7, the output current is shown as a function of the reference resistance.

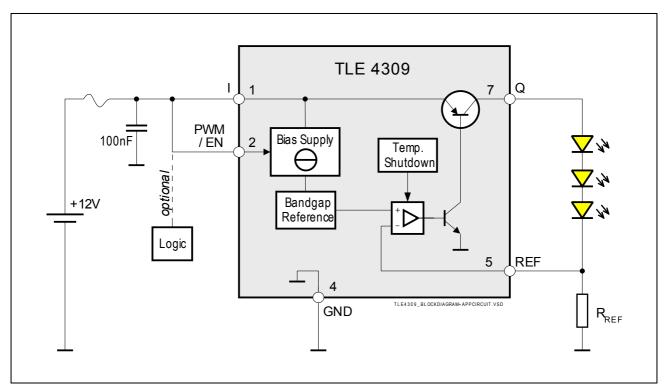


Figure 1 Block Diagram and Typical Application Circuit

During operation, the LED brightness is modifiable by adjusting the duty cycle at the PWM/EN input. This allows to dim the LEDs during operation or adapt the output current to different LED luminosity classes.

A low signal at the PWM/EN input sets the IC into sleep mode consuming less than 1 μ A. Due to its high input impedance, the PWM/EN pin can be used as an enable input. Connect the pin directly to the supply line, if its function is not needed.

The minimum supply voltage calculates as the sum of the LED forward voltages, the TLE4309 dropout voltage and the maximum voltage drop across the reference resistor $R_{\rm REF}$.



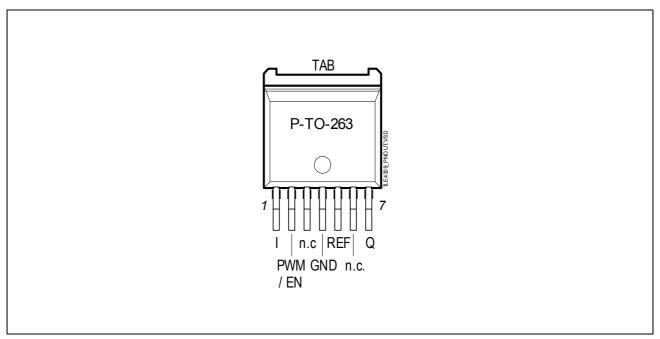


Figure 2 Pin Configuration (top view)

l able 1	Pin	Definitions and Functions
Pin No.	Symbol	Function
1	I	Input.
2	PWM / EN	Pulse Width Modulation Input / ENABLE input. If not needed connect to I (pin 1).
3	n. c.	Internally not connected. Leave open.
4	GND	Ground reference; connect to heatsink area and to TAB.
5	REF	Reference Input; connect to shunt resistor.
6	n. c.	Internally not connected. Leave open.
7	Q	Output.
TAB		Connect to heatsink area and to GND pin.

Table 1 Pin Definitions and Functions



Parameter	Symbol	Limit Values		Unit	Remarks
		Min.	Max.		
Input				•	
Voltage	$V_{\rm I}$	-42	45	V	-
Current	I	-	-	mA	internally limited
Output				•	
Voltage	V_{Q}	-1	40	V	-
Current	IQ	-	-	mA	internally limited
Reference Input					
Voltage	V_{REF}	-1	16	V	-
Current	I _{REF}	-2	2	mA	-
Pulse Width Modulatio		Input		•	
Voltage	V_{PWM}	-40	40	V	-
Current	I _{PWM}	-1	1	mA	-
Temperatures			·		
Junction temperature	Tj	-40	150	°C	-
Storage temperature	T _{stg}	-50	150	°C	-
ESD Susceptibility		•	·	•	
ESD Resistivity	$V_{\rm ESD,HBM}$	-2	2	kV	Human Body Model ¹⁾

Table 2Absolute Maximum Ratings

1) ESD susceptibility, HBM according to EIA/JESD 22-A114B

- Note: Stresses above the ones listed here may cause permanent damage to the device. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.
- Note: Integrated protection functions are designed to prevent IC destruction under fault conditions described in the data sheet. Fault conditions are considered as "outside" normal operating range. Protection functions are not designed for continuous repetitive operation.



Table 3Operating Range

Parameter	Symbol	Limit Values		Unit	Remarks
		Min.	Max.		
Input voltage	V_1	4.5	24	V	_
PWM / ENABLE voltage	V_{PWM}	0	24	V	_
Junction temperature	Tj	-40	150	°C	-
Reference resistor	R _{REF}	0	1.8	Ω	_

Note: Within the functional range the IC operates as described in the circuit description. The electrical characteristics are specified within the conditions given in the related electrical characteristics table

Table 4Thermal Resistance

Parameter	Symbol	Typical Limit Values	Unit	Remarks
Junction ambient	$R_{ m thja}$	78	K/W	Footprint only ¹⁾
		52	K/W	300mm ² heat sink area ¹⁾
		39	K/W	600mm ² heat sink area ¹⁾
Junction case	R _{thjc}	3	K/W	-

1) Worst case regarding peak temperature; mounted on PCB FR4, $80 \times 80 \times 1.5$ mm³, 35 µm Cu, horizontal position, zero airflow.



Table 5 Electrical Characteristics

 $V_{\rm I}$ = 13.5 V; $V_{\rm PWM} \ge V_{\rm PWM,H}$; -25 °C < $T_{\rm j}$ < 125 °C; all voltages with respect to ground; unless otherwise specified

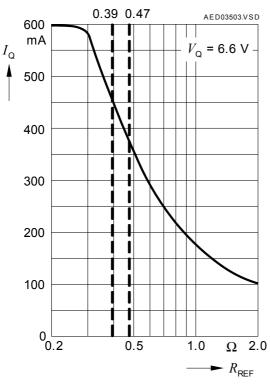
Parameter	Symbol	Limit Values			Unit	Test Condition
	-	Min.	Тур.	Max.		
Output	1					I
Output current	I _Q	171	178	185	mA	$(V_{\rm Q} - V_{\rm REF})^{1)} = 6.6 \text{ V};$ $R_{\rm REF} = 1.0 \Omega$
		438	456	474	mA	$(V_{\rm Q} - V_{\rm REF})^{1)} = 6.6 \text{ V};$ $R_{\rm REF} = 0.39 \Omega$
Output current limit	I _{Qmax}	-	600	-	mA	$R_{REF} = 0 \ \Omega$
Dropout voltage	V_{dr}	_	0.35	0.7	V	$I_{\rm Q} = 300 {\rm mA}$
Reference	•		1	1		
Reference Voltage	V_{REF}	172	178	184	mV	0.39 < $R_{\sf REF}$ < 1.0 Ω $T_{\sf j}$ = 25 °C
Reference Voltage	V_{REF}	171	178	185	mV	$0.39 < R_{\sf REF} < 1.0 \ \Omega$
Reference Input Current	I _{REF}	-1	0.1	1	μA	$V_{REF} = 180 \text{ mV}$
Current Consumptio	n					
Current consumption off mode	I _{q,OFF}	-	0.1	2	μA	PWM/EN = L, $T_{\rm j}$ < 85 °C
Current consumption	Iq	_	12	22	mA	$V_{\rm Q} = 6.6 \text{ V}; R_{\rm REF} = 0.47 \Omega$
PWM / ENABLE			1	1	1	
PWM high level valid	$V_{\rm PWM,H}$	2.6	_	_	V	-
PWM low level valid	$V_{PWM,L}$	_	_	0.7	V	-
PWM input current high level	I _{PWM,H}	-	220	500	μA	V _{PWM} = 5.0 V
PWM input current low level	I _{PWM,L}	-1	-	1	μA	V _{PWM} = 0.0 V
Turn on delay time	t _{PWM,ON}	0	15	40	μs	70% of I _{Qnom}
Turn off delay time	t _{PWM,OFF}	0	15	40	μs	30% of I _{Qnom}

1) $(V_Q - V_{REF})$ equals the forward voltage sum of the connected LEDs, see figure 1.



Typical Performance Characteristics

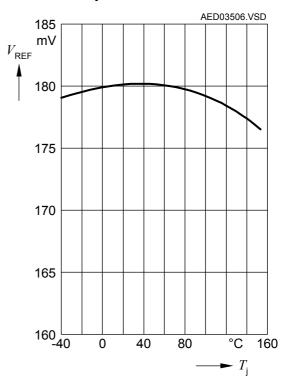
Output Current versus Reference Resistor



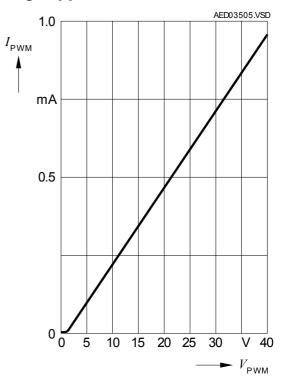
Output Current versus Supply Voltage



Reference Voltage versus Junction Temperature



PWM/EN Pin Input Current versus Voltage applied





Package Outline

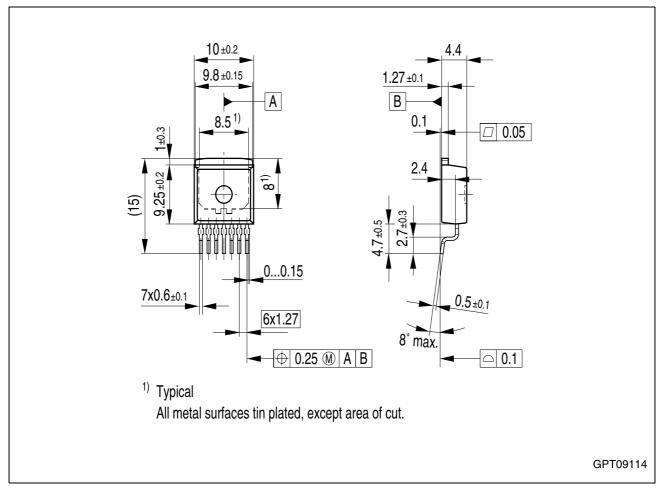


Figure 3 PG-TO263-7-1

Green Product (RoHS compliant)

To meet the world-wide customer requirements for environmentally friendly products and to be compliant with government regulations the device is available as a green product. Green products are RoHS-Compliant (i.e Pb-free finish on leads and suitable for Pb-free soldering according to IPC/JEDEC J-STD-020).

find all packages, sorts of packing and others on the Infineon Internet Page "Packages": http://www.infineon.com/packages.

SMD = Surface Mounted Device

Dimensions in mm



Application Note: Thermal Considerations ¹⁾

By describing an example, this section shows how the power dissipation and the needed thermal resistance can be estimated.

For a typical application circuit as shown in Figure 1, the following parameters shall apply:

- Number of LEDs in series: 3
- LED current desired: $I_{\rm F}$ = 260 mA
- LED minimum forward voltage: $V_{F,MIN} = 3.0 \text{ V}$
- LED maximum forward voltage: $V_{F,MAX} = 3.4 V$
- Supply minimum DC voltage: $V_{I,MIN} = 11.0 V$
- Supply maximum DC voltage: $V_{I,MAX} = 13.0 V$
- Maximum ambient temperature: $T_{a,max} = 85 \text{ °C}$

1. Selecting the reference resistor:

The reference resistor R_{REF} can be selected by using the formular on page 2 or by taking the value from the typical performance graph "Output Current versus Reference Resistor" on Page 7. In order to obtain a LED forward current of I_{F} = 260 mA, the graph shows an reference resistor of R_{REF} = 0.68 Ω .

2. Verifying the minimum supply voltage:

In order to be able to drive the LEDs with a constant current, the minimum supply voltage needed is:

$$V_{\rm I,MIN} = V_{\rm F,MAX,total} + V_{\rm dr} + V_{\rm REF}$$

whereas $V_{\text{F,MAX,total}}$ is the maximum forward voltage sum of the LEDs connected in series, V_{dr} the dropout voltage and V_{BFF} the reference voltage.

Since the power dissipation is low at a small voltage drop accross the regulator, $V_{dr} = 0.35$ V is assumed. The maximum reference voltags is $V_{REF} = 185$ mV. Therefore, a minimum supply voltage of 11 V is sufficient for the example given above.

3. Determining the minimum and maximum LED forward current:

The TLE4309 has an accuracy of < 4 %. With a reference resistor of R_{REF} = 0.68 Ω , the output current I_{Ω} will be between 251 mA and 272 mA.

¹⁾ This information is given as a hint for the implementation of the device only and shall not be regarded as a description or warranty of a certain functionality, condition or quality of the device.



4. Calculating the maximum power dissipation:

For calculating the maximum power dissipation, the minimum forward voltage sum of the LEDs connected in series $V_{\rm F,MIN,total}$ (= No. of LEDs * $V_{\rm F,MIN}$), the maximum supply voltage $V_{\rm I,MAX}$, the IC current consumption $I_{\rm q,MAX}$, as well as the the maximum output current $I_{\rm Q,MAX}$ needs to be considered. Hence, the maximum power dissipation $P_{\rm D,MAX}$ calculates:

 $P_{\rm D,MAX} = (V_{\rm I,MAX} - V_{\rm F,MIN,total}) * I_{\rm Q,MAX} + V_{\rm I,MAX} * I_{\rm q,MAX}$

With assuming a maximum current consumption of 15 mA @ I_Q = 260 mA, the maximum power dissipation for the example is 1.28 W.

5. Thermal Resistance needed:

The thermal resistance from junction to ambient $R_{\text{th.i-a}}$ calculates:

$$R_{\text{th,j-a}} = (T_{\text{j,max}} - T_{\text{a,max}}) / P_{\text{D,MAX}}$$

With allowing a junction temperature of 150 °C, the $R_{\text{th,j-a}}$ needed in our example would be 50 K/W.

In case some copper area on the PCB is used as a heat sink, the area needed is approximately 6 cm² (board in horizontal position, no airflow). If the area is not available, several via holes to the GND-layer or to a heatsink area on the PCB backside help to distribute the heat.

For additional information on the thermal resistance see Infineon's special subject book "Thermal Resistance - Theory and Practice" including extended package information.





Revision History

Version	Date	Changes
Rev. 1.0	2007-03-20	Final Datasheet and initial version of RoHS-compliant derivate of TLE4309G Page 1 and Page 8 : RoHS compliance statement and green product feature added Page 1 and Page 8 : Package changed to RoHS compliant version Legal Disclaimer updated

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