## \_General Description

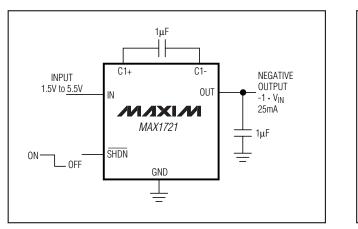
The ultra-small MAX1719/MAX1720/MAX1721 monolithic, CMOS charge-pump inverters accept input voltages ranging from +1.5V to +5.5V. The MAX1720 operates at 12kHz, and the MAX1719/MAX1721 operate at 125kHz. High efficiency, small external components, and logic-controlled shutdown make these devices ideal for both battery-powered and board-level voltage conversion applications.

Oscillator control circuitry and four power MOSFET switches are included on-chip. A typical MAX1719/ MAX1720/MAX1721 application is generating a -5V supply from a +5V logic supply to power analog circuitry. All three parts come in a 6-pin SOT23 package and can deliver a continuous 25mA output current.

For pin-compatible SOT23 switched-capacitor voltage inverters without shutdown (5-pin SOT23), see the MAX828/MAX829 and MAX870/MAX871 data sheets. For applications requiring more power, the MAX860/MAX861 deliver up to 50mA. For regulated outputs (up to -2 x VIN), refer to the MAX868. The MAX860/MAX861 and MAX868 are available in space-saving  $\mu$ MAX packages.

## \_ Applications

Local Negative Supply from a Positive Supply Small LCD Panels GaAs PA Bias Supply Handy-Terminals, PDAs Battery-Operated Equipment



## **Typical Operating Circuit**

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Maxim Integrated Products 1

For pricing, delivery, and ordering information, please contact Maxim/Dallas Direct! at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

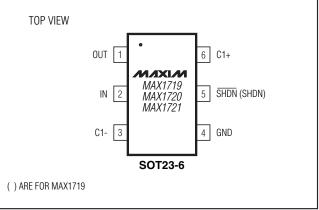
#### **Features**

- InA Logic-Controlled Shutdown
- 6-Pin SOT23 Package
- 99.9% Voltage Conversion Efficiency
- 50µA Quiescent Current (MAX1720)
- ♦ +1.5V to +5.5V Input Voltage Range
- ♦ 25mA Output Current
- Requires Only Two 1µF Capacitors (MAX1719/MAX1721)

### **Ordering Information**

PART	TEMP RANGE	PIN- PACKAGE	SOT TOP MARK
MAX1719EUT	-40°C to +85°C	6 SOT23-6	AACA
MAX1720EUT	-40°C to +85°C	6 SOT23-6	AABS
MAX1721EUT	-40°C to +85°C	6 SOT23-6	AABT

### **Pin Configuration**



### **ABSOLUTE MAXIMUM RATINGS**

IN to GND	-0.3V to +6V
OUT to GND	-6V to +0.3V
C1+, SHDN, SHDN to GND	0.3V to (V <sub>IN</sub> + 0.3V)
C1- to GND	(V <sub>OUT</sub> - 0.3V) to +0.3V
OUT Output Current	
OUT Short Circuit to GND	

Continuous Power Dissipation ( $T_A = +70^{\circ}C$ )	
6-Pin SOT23 (derate 8.7mW/°C above +70°	°C)696mW
Operating Temperature Range	40°C to +85°C
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## **ELECTRICAL CHARACTERISTICS**

 $(V_{IN} = +5V, SHDN = GND (MAX1719), \overline{SHDN} = IN (MAX1720/MAX1721), C1 = C2 = 10\mu F (MAX1720), C1 = C2 = 1\mu F (MAX1719/MAX1721), circuit of Figure 1, TA = 0°C to +85°C, unless otherwise noted. Typical values are at TA = +25°C.)$ 

PARAMETER	SYMBOL	CONDITIONS		MIN	ТҮР	MAX	UNITS	
		MAX1720	$T_A = +25^{\circ}C$	1.25		5.5	- V	
	Vin	$R_L = 10k\Omega$	$T_A = 0^\circ C \text{ to } + 85^\circ C$	1.5		5.5		
Supply Voltage Range	VIIN	MAX1719/MAX1721	$T_A = +25^{\circ}C$	1.4		5.5		
		$R_L = 10k\Omega$	$T_A = 0^{\circ}C \text{ to } + 85^{\circ}C$	1.5		5.5		
Quiescent Supply Current	lcc	TA = +25°C	MAX1720		50	90	μA	
(Note 3)		TA = +23 C	MAX1719/MAX1721		350	650		
Shutdown Supply Current	ISHDN	$\frac{\text{SHDN} = \text{IN} (\text{MAX1719})}{\text{SHDN} = \text{GND}}$	$T_A = +25^{\circ}C$		0.001	1	μΑ	
	ISHDIN	(MAX1720/MAX1721)	$T_A = +85^{\circ}C$		0.02			
Oscillator Frequency	fooo	$T_A = +25^{\circ}C$	MAX1720	7	12	17	- kHz	
	fosc		MAX1719/MAX1721	70	125	180		
Voltage Conversion Efficiency		Iout = 0, TA = +25°C		99	99.9		%	
	R <sub>O</sub>	I <sub>OUT</sub> = 10mA	$T_A = +25^{\circ}C$		23	50	Ω	
Output Resistance (Note 1)			$T_A = 0^{\circ}C \text{ to } +85^{\circ}C$			65		
OUT to GND Shutdown Resistance	Ro, SHDN	SHDN = IN (MAX1719), SHDN = GND (MAX1720/MAX1721), OUT is internally forced to GND in shutdown			4	12	Ω	
SHDN/ SHDN Input Logic High	Vih	$+2.5V \le V_{\rm IN} \le +5.5V$		2.0			V	
	VIH	$V_{\rm IN (MIN)} \le V_{\rm IN} \le +2.5V$		V <sub>IN</sub> - 0.2			v	
SHDN/ SHDN Input Logic Low	VIL	$+2.5V \le V_{IN} \le +5.5V$				0.6	V	
Shibh Shibh hiput Eogle Eow	VIL	$V_{IN}$ (MIN) $\leq V_{IN} \leq +2.5V$				0.2	v	
SHDN/SHDN Bias Current	I <sub>IL</sub> , I <sub>IH</sub>	$SHDN/\overline{SHDN} = GND$	$T_A = +25^{\circ}C$	-100	0.05	100	nA	
		or VIN	$T_A = +85^{\circ}C$		10			
Wake-Up Time from Shutdown		lоuт = 5mA	MAX1720		800		– μs	
			MAX1719/MAX1721		80			

## **ELECTRICAL CHARACTERISTICS**

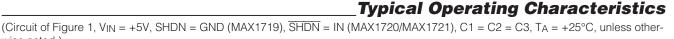
 $(V_{IN} = +5V, SHDN = GND (MAX1719), \overline{SHDN} = IN (MAX1720/MAX1721), C1 = C2 = 10\mu F (MAX1720), C1 = C2 = 1\mu F (MAX1719/MAX1721), circuit of Figure 1, TA = -40°C to +85°C, unless otherwise noted.) (Note 2)$ 

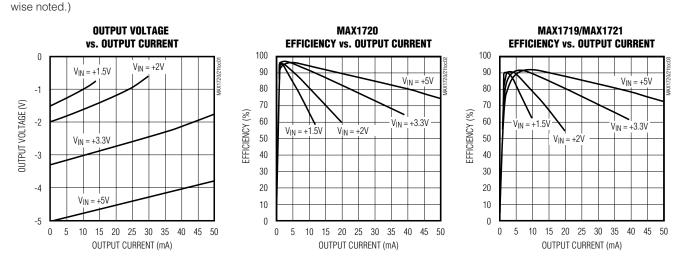
PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
		$R_L = 10k\Omega$	MAX1720	1.5		5.5	V
Supply Voltage Range	VIN		MAX1719/MAX1721	1.6		5.5	
Quiescent Current (Note 3)	lee	MAX1720				100	μΑ
Quescent Current (Note 3)	Icc	MAX1719/MAX1721				750	
Oscillator Frequency	fooo	MAX1720		6		21	kHz
Oscillator Frequency	fosc	MAX1719/MAX1721		60		200	
Voltage Conversion Efficiency		I <sub>OUT</sub> = 0		99			%
Output Resistance (Note 1)	Ro	I <sub>OUT</sub> = 10mA				65	Ω
Output Current	lout	Continuous, long-term				25	mA <sub>RMS</sub>
OUT to GND Shutdown Resistance	Ro, SHDN	SHDN = IN (MAX1719), SHDN = GND (MAX1720/MAX1721), OUT is internally forced to GND in shutdown				12	Ω
CHDN/CHDN Input Logia High	Vuu	$+2.5V \le V_{IN} \le +5.5V$		2.1			V
SHDN/ SHDN Input Logic High	ViH	$V_{IN (MIN)} \le V_{IN} \le +2.5V$		V <sub>IN</sub> - 0.2			1
	VIL	$+2.5V \le V_{IN} \le +5.5V$				0.6	- V
SHDN/ SHDN Input Logic Low	VIL	$V_{IN (MIN)} \le V_{IN} \le +2.5V$				0.2	

Note 1: Capacitor contribution (ESR component plus (1/fOSC) · C) is approximately 20% of output impedance.

Note 2: All specifications from -40°C to +85°C are guaranteed by design, not production tested.

**Note 3:** The MAX1719/MAX1720/MAX1721 may draw high supply current during startup, up to the minimum operating supply voltage. To guarantee proper startup, the input supply must be capable of delivering 90mA more than the maximum load current.

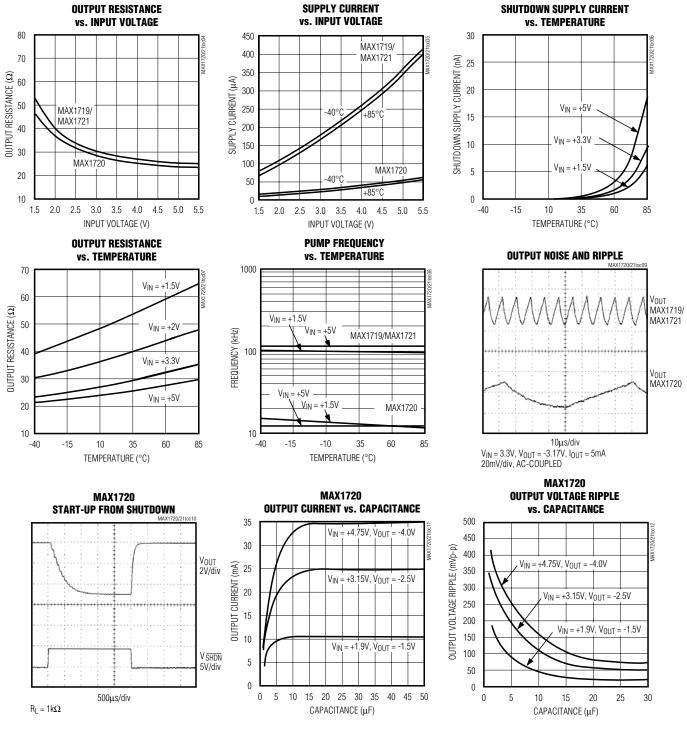




### **Typical Operating Characteristics (continued)**

MIXIM

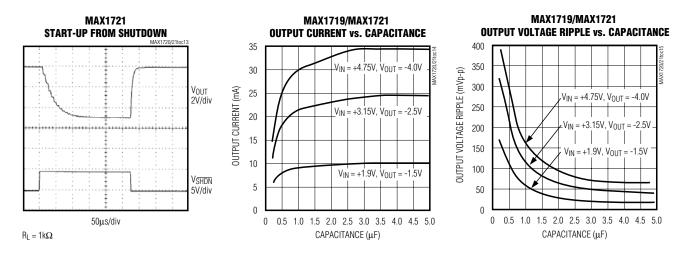
(Circuit of Figure 1,  $V_{IN} = +5V$ , SHDN = GND (MAX1719),  $\overline{SHDN} = IN$  (MAX1720/MAX1721), C1 = C2 = C3, T<sub>A</sub> = +25°C, unless otherwise noted.)



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### \_Typical Operating Characteristics (continued)

(Circuit of Figure 1,  $V_{IN} = +5V$ , SHDN = GND (MAX1719),  $\overline{SHDN} = IN$  (MAX1720/MAX1721), C1 = C2 = C3, T<sub>A</sub> = +25°C, unless otherwise noted.)



## **Pin Description**

Р	IN		FUNCTION	
MAX1719	MAX1720 MAX1721	NAME		
1	1	OUT	Inverting Charge-Pump Output	
2	2	IN	Power-Supply Positive Voltage Input	
3	3	C1-	Negative Terminal of Flying Capacitor	
4	4	GND	Ground	
5	-	SHDN	Noninverting Shutdown Input. Drive this pin low for normal operation; drive it high for shutdown mode. <b>OUT is actively pulled to ground during shutdown.</b>	
_	5	SHDN	Inverting Shutdown Input. Drive this pin high for normal operation; drive it low for shutdown mode. <b>OUT is actively pulled to ground during shutdown.</b>	
6	6	C1+	Positive Terminal of Flying Capacitor	

## **Detailed Description**

The MAX1719/MAX1720/MAX1721 capacitive charge pumps invert the voltage applied to their input. For highest performance, use low equivalent series resistance (ESR) capacitors (e.g., ceramic).

During the first half-cycle, switches S2 and S4 open, switches S1 and S3 close, and capacitor C1 charges to the voltage at IN (Figure 2). During the second half-

cycle, S1 and S3 open, S2 and S4 close, and C1 is level shifted downward by V<sub>IN</sub> volts. This connects C1 in parallel with the reservoir capacitor C2. If the voltage across C2 is smaller than the voltage across C1, charge flows from C1 to C2 until the voltage across C2 reaches -V<sub>IN</sub>. The actual voltage at the output is more positive than -V<sub>IN</sub>, since switches S1–S4 have resistance and the load drains charge from C2.

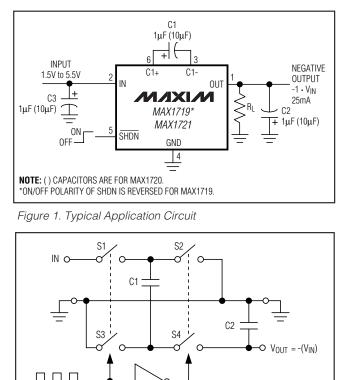


Figure 2. Ideal Voltage Inverter

#### **Charge-Pump Output**

The MAX1719/MAX1720/MAX1721 are not voltage regulators: the charge pumps' output resistance is approximately  $23\Omega$  at room temperature (with V<sub>IN</sub> = +5V), and V<sub>OUT</sub> approaches -5V when lightly loaded. V<sub>OUT</sub> will droop toward GND as load current increases. The droop of the negative supply (V<sub>DROOP</sub>-) equals the current draw from OUT (I<sub>OUT</sub>) times the negative converter's output resistance (R<sub>O</sub>):

VDROOP- = IOUT x RO

The negative output voltage will be:

VOUT = -(VIN - VDROOP)

#### **Efficiency Considerations**

The efficiency of the MAX1719/MAX1720/MAX1721 is dominated by its quiescent supply current (Iq) at low output current and by its output impedance (R<sub>OUT</sub>) at higher output current; it is given by:

$$\eta \cong \frac{I_{OUT}}{I_{OUT} + I_{Q}} \left( 1 - \frac{I_{OUT} \times R_{OUT}}{V_{IN}} \right)$$

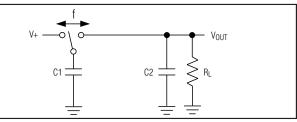


Figure 3a. Switched-Capacitor Model

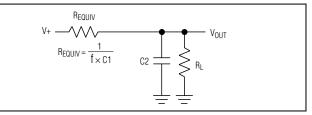


Figure 3b. Equivalent Circuit

where the output impedance is roughly approximated by:

$$R_{OUT} \cong \frac{1}{(f_{OSC}) \times C1} + 2R_{SW} + 4ESR_{C1} + ESR_{C2}$$

The first term is the effective resistance of an ideal switched-capacitor circuit (Figures 3a and 3b), and R<sub>SW</sub> is the sum of the charge pump's internal switch resistances (typically  $8\Omega$  to  $9\Omega$  at V<sub>IN</sub> = +5V). The typical output impedance is more accurately determined from the *Typical Operating Characteristics*.

#### Shutdown Mode

The MAX1719/MAX1720/MAX1721 have a logic-controlled shutdown input. Driving SHDN low places the MAX1720/MAX1721 in a low-power shutdown mode. The MAX1719's shutdown input is inverted from that of the MAX1720/MAX1721. Driving SHDN high places the MAX1719 in a low-power shutdown mode. The charge-pump switching halts, supply current is reduced to 1nA, and OUT is actively pulled to ground through a 4 $\Omega$  resistance.

### **Applications Information**

#### **Capacitor Selection**

To maintain the lowest output resistance, use capacitors with low ESR (Table 1). The charge-pump output resistance is a function of C1's and C2's ESR. Therefore, minimizing the charge-pump capacitor's ESR minimizes the total output resistance. Table 2 gives suggested capacitor values for minimizing output resistance or minimizing capacitor size.



#### Flying Capacitor (C1)

Increasing the flying capacitor's value reduces the output resistance. Above a certain point, increasing C1's capacitance has a negligible effect because the output resistance becomes dominated by the internal switch resistance and capacitor ESR.

#### **Output Capacitor (C2)**

Increasing the output capacitor's value reduces the output ripple voltage. Decreasing its ESR reduces both output resistance and ripple. Lower capacitance values can be used with light loads if higher output ripple can be tolerated. Use the following equation to calculate the peak-to-peak ripple:

$$V_{\text{RIPPLE}} = \frac{I_{\text{OUT}}}{2 \times f_{\text{OSC}} \times C2} + 2 \times I_{\text{OUT}} \times \text{ESR}_{C2}$$

#### Input Bypass Capacitor (C3)

Bypass the incoming supply to reduce its AC impedance and the impact of the MAX1719/MAX1720/MAX1721's switching noise. A bypass capacitor with a value equal to that of C1 is recommended.

#### Voltage Inverter

The most common application for these devices is a charge-pump voltage inverter (Figure 1). This application requires only two external components-capacitors C1 and C2-plus a bypass capacitor, if necessary. Refer to the Capacitor Selection section for suggested capacitor types.

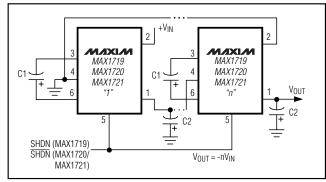


Figure 4. Cascading MAX1719s or MAX1720s or MAX1721s to Increase Output Voltage

#### **Cascading Devices**

Two devices can be cascaded to produce an even larger negative voltage (Figure 4). The unloaded output voltage is normally -2 x VIN, but this is reduced slightly by the output resistance of the first device multiplied by the quiescent current of the second. When cascading more than two devices, the output resistance rises dramatically. For applications requiring larger negative voltages, see the MAX865 and MAX868 data sheets. The maximum load current and startup current of the n<sup>th</sup> cascaded circuit must not exceed the maximum output current capability of the (n - 1)th circuit to ensure proper startup.

PRODUCTION METHOD	MANUFACTURER	SERIES	PHONE	FAX
	AVX	TPS series	803-946-0690	803-626-3123
Surface-Mount Tantalum	Matsuo	267 series	714-969-2491	714-960-6492
	Sprague	593D, 595D series	603-224-1961	603-224-1430
Surface-Mount	AVX	X7R	803-946-0690	803-626-3123
Ceramic	Matsuo	X7R	714-969-2491	714-960-6492

#### Table 1. Low-ESR Capacitor Manufacturers

MXXIM

### Table 2. Capacitor Selection for Minimum Output Resistance or Capacitor Size

PART	fosc	CAPACITORS TO MINIMIZE OUTPUT RESISTANCE ( $R_0 = 23\Omega$ , TYP) C1 = C2	CAPACITORS TO MINIMIZE SIZE ( $R_0 = 40\Omega$ , TYP) C1 = C2
MAX1720	12kHz	10µF	3.3µF
MAX1719/MAX1721	125kHz	1µF	0.33µF

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MAX1719/MAX1720/MAX1721

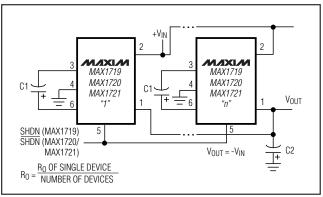


Figure 5. Paralleling MAX1719s or MAX1720s or MAX1721s to Reduce Output Resistance

#### **Paralleling Devices**

Paralleling multiple MAX1719s, MAX1720s, or MAX1721s reduces the output resistance. Each device requires its own pump capacitor (C1), but the reservoir capacitor (C2) serves all devices (Figure 5). Increase C2's value by a factor of n, where n is the number of parallel devices. Figure 5 shows the equation for calculating output resistance.

#### **Combined Doubler/Inverter**

In the circuit of Figure 6, capacitors C1 and C2 form the inverter, while C3 and C4 form the doubler. C1 and C3 are the pump capacitors; C2 and C4 are the reservoir capacitors. Because both the inverter and doubler use part of the charge-pump circuit, loading either output causes both outputs to decline toward GND. Make sure the sum of the currents drawn from the two outputs does not exceed 25mA.

#### Heavy Load Connected to a Positive Supply

Under heavy loads, where a higher supply is sourcing current into OUT, the OUT supply must not be pulled above ground. Applications that sink heavy current into

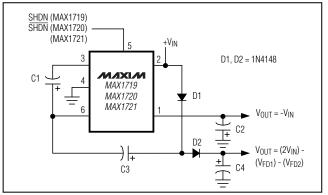


Figure 6. Combined Doubler and Inverter

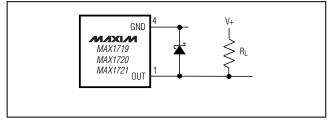


Figure 7. Heavy Load Connected to a Positive Supply

OUT require a Schottky diode (1N5817) between GND and OUT, with the anode connected to OUT (Figure 7).

#### Layout and Grounding

Good layout is important, primarily for good noise performance. To ensure good layout, mount all components as close together as possible, keep traces short to minimize parasitic inductance and capacitance, and use a ground plane.

**Chip Information** 

TRANSISTOR COUNT: 85

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