



3-Port 10/100 Integrated Switch with PHY and Frame Buffer

Rev. 2.06

General Description

The KS8993 contains three 10/100 physical layer transceivers, three MAC (Media Access Control) units with an integrated layer 2 switch. The device runs in two modes. The first mode is a three port integrated switch and the second is as a three port switch with the third port decoupled from the physical port. In this mode access to the third MAC is provided using a reverse or forward MII (Media Independent Interface) such that an external MAC can be directly connected to the KS8993. This interface also supports the 7-wire (serial network interface) as used by some routing devices.

Useful configurations include a stand alone three port switch as well as a two port switch with a routing element connected to the extra MII port. The additional port is also useful for public network interfacing.

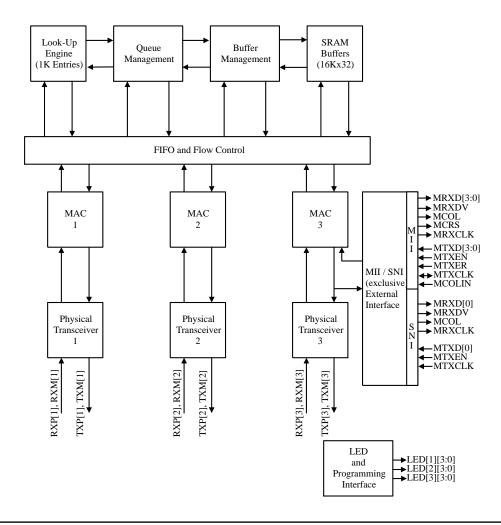
The KS8993 has rich features such as VLAN and priority queuing and is designed to reside in an unmanaged design not requiring processor intervention. This is achieved through I/O strapping at system reset time.

On the media side, the KS8993 supports 10BaseT, 100BaseTX and 100BaseFX as specified by the IEEE 802.3 committee.

Physical signal transmission and reception are enhanced through use of analog circuitry that makes the design more efficient and allows for lower power consumption and smaller chip die size.

Data sheets and support documentation can be found on Micrel's web site at www.micrel.com.

Functional Diagram



Features

- 3-port 10/100 integrated switch with physical layer transceivers
- · 64k Byte of SRAM on chip for frame buffering
- 1.0Gbps high performance memory bandwidth
- 10BaseT, 100BaseTX and 100BaseFX modes of operation
- Support for UTP or fiber installations
- Superior analog technology for reduced power and die size
- Supports port based VLAN
- QoS feature!! Supports 802.1p based priority or portbased priority
- Indicators for link, activity, full/half-duplex and speed
- Unmanaged operation via strapping at system reset time
- Hardware based 10/100, full/half, flow control and autonegotiation
- Individual port forced modes (full-duplex, 100BaseTX) when auto-negotiation is disabled
- Wire speed reception and transmission
- On chip integrated address look-up engine, supports 1K absolute MAC addresses
- Automatic address learning, address aging and address migration
- Full-duplex IEEE 802.3x flow control (Pause) with force mode option
- · Half-duplex back pressure flow control
- Comprehensive LED support
- External MAC interface (MII or SNI 7-wire) for router applications
- 300mA (0.75W) including physical transmit drivers
- Commercial temperature range: 0°C to +70°C
- Available in 128-pin PQFP with single 2.5V power supply

Ordering Information

Part Numbe	r	Temperature	Dankana	
Standard	Pb-Free	Range	Package	
KS8993	KSZ8993	0°C to +70°C	128-Pin PQFP	

Revision History

Revision	Date	Summary of Changes
1.00	04/13/00	Document origination
1.01	05/31/00	Miscellaneous changes
1.02	06/08/00	Index repair
1.03	09/20/00	MII forward correction. MRXD[3:1] correction.
1.04	10/30/00	Update voltage ratings.
1.05	10/31/00	Correct I/O descriptions.
1.06	11/08/00	Correct mode operation for LED[1:3][0] Add timing information
1.07	12/21/00	Correct pin information
1.08	03/23/01	Correct VLAN description.
1.09	03/26/01	Update MODESEL descriptions for packet size extensions
1.10	04/19/01	Update electrical characteristics; Correct I/O information.
1.11	04/20/01	Correct timing information
1.12	05/10/01	Update I/O descriptions
1.13	06/08/01	Define control for LED[3][3]
1.14	06/26/01	Revise definition for LED[3][3]
1.15	08/1/01	Update timing information and power dissipation Add power up timing description; Correct DISAN3 default mode.
1.16	08/9/01	Correct LED [1] [1] to float configuration Add Reverse and Forward MII timing
2.00	4/8/02	Correct reserve buffer from 128 to 96 for PRSV pin. Add max. current. Add force flow control Option as follows: Change pin 50 from reserved to FFLOW1# for force flow control on port 1.) Change pin 46 from reserved to FFLOW2# for force flow control on port 2. Modify LED[1][2] for force flow control on port 3.
2.01	5/6/02	Add TX Disable for Port 1 and port 2, Power down for port 3 and Far end Fault Disable features using MUX[1:2] and TEST[1:2] pins.
2.02	7/2/02	Recommend pull-down on LED[3][3]
2.03	8/29/03	Convert to new format.
2.04	1/24/05	Added reset circuit recommendation.
2.05	5/12/05	Added lead-free part number

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System Level Applications

The KS8993 can be configured to fit either in a three port 10/100 application or as a two port 10/100 network interface with an extra MII or SNI port. This MII/SNI port can be connected to an external processor and used for routing purposes or

public network access. The major benefits of using the KS8993 are the lower power consumption, unmanaged operation, flexible configuration and built in frame buffering. Two such applications are depicted below.

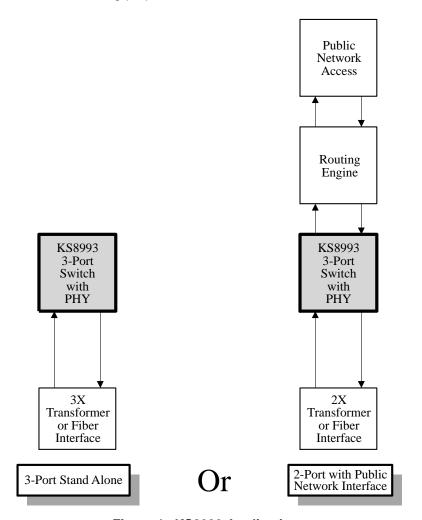


Figure 1. KS8993 Applications

Pin Description

Pin Number	Pin Name	Type ^(Note 1)	Port	Pin Function		
1	GND_ANA	GND		Analog ground		
2	MUX[2]	I		Factory test pin		
3	MUX[1]	I		Factory test pin		
4	GND_RX[1]	GND	1	Ground for receiver		
5	RXP[1]	I	1	Physical receive signal + (differential)		
6	RXM[1]	I	1	Physical receive signal - (differential)		
7	VDD_RX[1]	Pwr	1	2.5V for receiver		
8	VREF[1]	0	1	Reference voltage for transmit transformer center tap		
9	TXP[1]	0	1	Physical transmit signal + (differential)		
10	TXM[1]	0	1	Physical transmit signal - (differential)		
11	GND_TX[1]	GND	1	Ground for transmit circuitry		
12	VDD_TX[1]	Pwr	1	2.5V for transmit circuitry		
13	VDD_BG	Pwr		2.5V for analog circuitry		
14	ISET	0		Set physical transmit output current		
15	GND_BG	GND		Ground for analog circuitry		
16	GND_PLL	GND		Ground for phase locked loop circuitry		
17	VDD_PLL	Pwr		2.5V for phase locked loop circuitry		
18	GND_RX[2]	GND	2	Ground for receiver		
19	RXP[2]	I	2	Physical receive signal + (differential)		
20	RXM[2]	I	2	Physical receive signal - (differential)		
21	VDD_RX[2]	Pwr	2	2.5V for receiver		
22	VREF[2]	0	2	Reference voltage for transmit transformer center tap		
23	TXP[2]	0	2	Physical transmit signal + (differential)		
24	TXM[2]	0	2	Physical transmit signal - (differential)		
25	GND_TX[2]	GND	2	Ground for transmit circuitry		
26	VDD_TX[2]	Pwr	2	2.5V for transmit circuitry		
27	VDD_TX[3]	Pwr	3	2.5V for transmit circuitry		
28	GND_TX[3]	GND	3	Ground for transmit circuitry		
29	TXP[3]	0	3	Physical transmit signal + (differential)		
30	TXM[3]	0	3	Physical transmit signal - (differential)		
31	VREF[3]	0	3	Reference voltage for transmit transformer center tap		
32	VDD_RX[3]	Pwr	3	2.5V for receiver		
33	RXP[3]	I	3	Physical receive signal + (differential)		
34	RXM[3]	I	3	Physical receive signal - (differential)		
35	GND_RX[3]	GND	3	Ground for receiver		
36	FXSD[2]	I	2	Fiber signal detect		
37	FXSD[3]	I	3	Fiber signal detect		
38	GND_ANA	GND		Analog ground		

Note 1. Pwr = power supply

GND = ground

I = input

O = output

I/O = bi-directional

Pin Number	Pin Name	Type ^(Note 1)	Port	Pin Function		
39	TEST[1]	I		Factory test pin		
40	TEST[2]	I		Factory test pin		
41	GND_RCV[2]	GND	2	Ground for clock recovery circuitry		
42	VDD_RCV[2]	Pwr	2	2.5V for clock recovery circuitry		
43	GND_RCV[3]	GND	3	Ground for clock recovery circuitry		
44	VDD_RCV[3]	Р	3	2.5V for clock recovery circuitry		
45	VMDIS	I		DIScard VLAN Mismatch packets		
46	FFLOW2#	I	2	Force flow control on port 2		
47	PV32	I	3	Port 3 VLAN Port mask bit 1		
48	PV31	I	3	Port 3 VLAN Port mask bit 0		
49	PV23	I	2	Port 2 VLAN Port mask bit 2		
50	FFLOW1#	I	1	Force flow control on port 1		
51	PV21	Į.	2	Port 2 VLAN Port mask bit 0		
52	PV13	I	1	Port 1 VLAN Port mask bit 2		
53	PV12	I	1	Port 1 VLAN Port mask bit 1		
54	DISAN3	I	3	Port 3 auto-negotiation disable (pull this down to enable port 3 auto negotiation)		
55	VDD	Pwr		2.5V for core digital circuitry		
56	GND	GND		Ground for digital circuitry		
57	MTXEN	I	3	MII transmit enable		
58	MTXD[3]	I	3	MII transmit bit 3		
59	MTXD[2]	I	3	MII transmit bit 2		
60	MTXD[1]	I	3	MII transmit bit 1		
61	MTXD[0]	I	3	MII transmit bit 0		
62	MTXER	I	3	MII transmit error		
63	MTXCLK	I/O	3	MII output clock		
64	MRXDV	0	3	MII receive data valid		
65	MRXD[3]	0	3	MII receive bit 3		
66	MRXD[2]	0	3	MII receive bit 2		
67	MRXD[1]	0	3	MII receive bit 1		
68	MRXD[0]	0	3	MII receive bit 0		
69	VDD_IO	Pwr		2.5V or 3.3V for MII interface, LEDs and other digital I/O		
70	GND	GND		Ground for digital circuitry		
71	MRXCLK	I/O	3	MII input clock		
72	MCOL	0	3	MII collision detect output		
73	MCRS	I/O	3	MII carrier sense		
74	MCOLIN	I	3	MII collision detect input		
75	MIIS[1]	I	3	MII mode select bit 1		
76	MIIS[0]	I	3	MII mode select bit 0		

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Pin Number	Pin Name	Type ^(Note 1)	Port	Pin Function	
77	MODESEL[3]	I		Selects LED and test modes	
78	MODESEL[2]	I		Selects LED and test modes	
79	MODESEL[1]	I		Selects LED and test modes	
80	MODESEL[0]	I		Selects LED and test modes	
81	TESTEN	I		Factory test pin - tie low for normal operation	
82	SCANEN	I		Factory test pin - tie low for normal operation	
83	RST#	I		Reset	
84	VDD	Pwr		2.5V for core digital circuitry	
85	GND	GND		Ground for digital circuitry	
86	LED[1][3]	0	1	Port 1 LED indicator 3	
87	LED[1][2]	0	1	Port 1 LED indicator 2	
88	LED[1][1]	0	1	Port 1 LED indicator 1	
89	LED[1][0]	0	1	Port 1 LED indicator 0	
90	LED[2][3]	0	2	Port 2 LED indicator 3	
91	LED[2][2]	0	2	Port 2 LED indicator 2	
92	LED[2][1]	0	2	Port 2 LED indicator 1	
93	LED[2][0]	0	2	Port 2 LED indicator 0	
94	VDD_IO	Pwr		2.5V or 3.3V for MII interface, LEDs and other digital I/O	
95	GND	GND		Ground for digital circuitry	
96	LED[3][3]	0	3	Port 3 LED indicator 3	
97	LED[3][2]	0	3	Port 3 LED indicator 2	
98	LED[3][1]	0	3	Port 3 LED indicator 1	
99	LED[3][0]	0	3	Port 3 LED indicator 0	
100	PRSV	I		Priority queue buffer reserve	
101	PRSEL[1]	I		Priority scheme select bit 1	
102	PRSEL[0]	I		Priority scheme select bit 0	
103	PBASE2	I		Priority base value bit 2	
104	PBASE1	I		Priority base value bit 1	
105	PBASE0	I		Priority base value bit 0	
106	P3_1PEN	I	3	Port 3 802.1p receive priority classification enable	
107	P2_1PEN	I	2	Port 2 802.1p receive priority classification enable	
108	P1_1PEN	I	1	Port 1 802.1p receive priority classification enable	
109	P3_TXQ2	I	3	Port 3 transmit queue split, priority queueing enable	
110	P2_TXQ2	I	2	Port 2 transmit queue split, priority queueing enable	
111	P1_TXQ2	I	1	Port 1 transmit queue split, priority queueing enable	
112	GND	GND		Ground for digital circuitry	
113	VDD	Pwr		2.5V for core digital circuitry	
114	P3_PP	I	3	Port 3 receive port based priority classification	
115	P2_PP	I	2	Port 2 receive port based priority classification	

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Pin Number	Pin Name	Type ^(Note 1)	Port	Pin Function		
116	P1_PP	I	1	Port 1 receive port based priority classification		
117	P1_TAGINS	I	1	Port 1 tag insertion enable		
118	P2_TAGINS	I	2	Port 2 tag insertion enable		
119	P3_TAGINS	I	3	Port 3 tag insertion enable		
120	P3_TAGRM	I	3	Port 3 tag removal enable		
121	P2_TAGRM	I	2	Port 2 tag removal enable		
122	P1_TAGRM	I	1	Port 1 tag removal enable		
123	VDD_RCV[1]	Pwr	1	2.5V for clock recovery circuitry		
124	GND_RCV[1]	GND	1	Ground for clock recovery circuitry		
125	X2	0		Connect to crystal input		
126	X1	I		Crystal or clock input		
127	FXSD[1]	Ī	1	Fiber signal detect		
128	AOUT	0		Factory test output		

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I/O Grouping

Description
Physical Interface
Media Independant Interface
Serial Network Interface
LED Indicators
Unmanaged Programmable
Control and Miscellaneous
Test (Factory)
Power and Ground

I/O Descriptions

Group	I/O Names	Active Status	Description
PHY	RXP[1:3] RXM[1:3]	Analog	Differential inputs (receive) for connection to media (transformer or fiber module).
	TXP[1:3] TXM[1:3]	Analog	Differential outputs (transmit) for connection to media (transformer or fiber module).
	FXSD[1:3]	Н	Fiber signal detect - connect to fiber signal detect output on fiber module. Tie low for 100TX mode.
	VREF[1:3]	Analog	Center tap transformer reference for transmit data.
	ISET	Analog	Transmit Current Set. Connecting an external reference resistor to set transmitter output current. This pin connects a 1% $3k\Omega$ resistor if a transformer of turns ratio of 1:1 is used.
MII			See "Table 2, MII Interconnect" for forward and reverse signal usage.
	MRXD[0:3]	Н	Four bit wide data bus for receiving MAC frames.
	MRXDV	Н	Receive data valid.
	MCRS	Н	Receive carrier sense.
	MCOL	Н	Receive collision detection.
	MCOLIN	Н	Collision in (for forward operation only).
	MRXCLK	Clock	Receive clock.
	MTXD[0:3]	Н	Four bit wide data bus for transmitting MAC frames.
	MTXEN	Н	Transmit enable.
	MTXER	Н	Transmit error.
	MTXCLK	Clock	Transmit clock.
SNI	MTXD[0]	Н	Serial transmit data.
	MTXEN	Н	Transmit enable.
	MTXCLK	Clock	Transmit clock.
	MRXD[0]	Н	Serial receive data.
	MRXDV	Н	Receive carrier sense/data valid.
	MCOL	Н	Collision detection.
	MRXCLK	Clock	Receive clock.
IND	LED[1:3][0]	L	Output (after reset). Mode 0: Speed (low = 100/high = 10). Mode 1: Reserved. Mode 2: Collision (toggle = collision during receiving, high = no collision). Mode 3: Speed (low = 100/high = 10).
	LED[1:3][1]	L	Output (after reset). Mode 0: Duplex (low = full/high = half). Mode 1: Duplex (low = full/high = half). Mode 2: Duplex (low = full/high = half). Mode 3: Reserved.
	LED[1:3][2]	L	Output (after reset). Mode 0: Collision (toggle = collision during receiving , high = no collision). Mode 1: Transmit Activity (toggle during transmission, high = idle). Mode 2: 10/link/act (constant low = link, toggle = act, constant high = no link). Mode 3: Full-Duplex + Collision (constant low = full-duplex, toggle = collision in half. duplex, constant high = half-duplex with no collision).

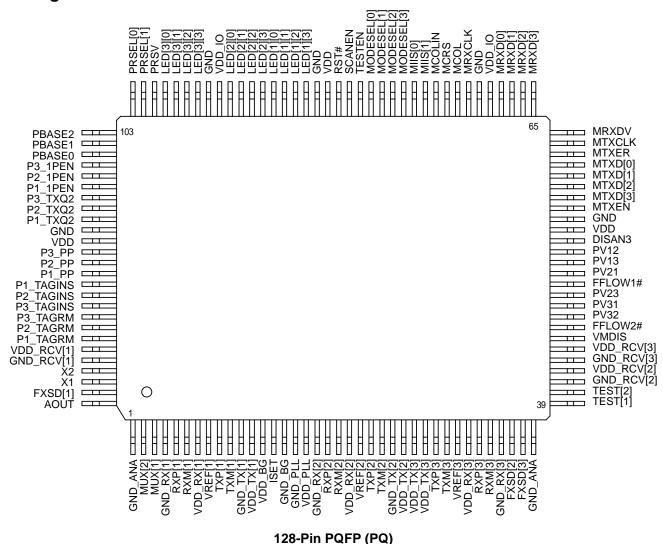
Group	I/O Names	Active Status	Desc	Description							
UP	LED[1:3][3] MODESEL[3:0]	T	Output (after reset). Mode 0: Link + Activity (toggle = receiving or transmitting, constant low = link, constant high = no link). Mode 1: Receive Activity (toggle during receiving / high = no receiving activity). Mode 2: 100/link/act (constant low = link, toggle = act, constant high = no link). Mode 3: Mode 3: Link + Activity (toggle = receiving or transmitting, constant low = link, constant high = no link). Note: Mode is set by MODESEL[3:0]; please see description in UP "Unmanaged Programming" section. Mode select at reset time. LED mode is selected by using the table below. MODESEL								
			also d	ontrols t			ne length accepted.				
			MOD	ESEL	,						
			3	2	1	0	LED mode	Max Length (no tag/tag)			
			0	0	0	0	LED mode 0	1518/1522			
			0	0	0	1	LED mode 1	1518/1522			
			0	0	1	0	LED mode 2	1518/1522			
			0	0	1	1	LED mode 3	1518/1522			
			0	1	0	0	Factory testing	Not applicable			
			0	1	0	1	Factory testing	Not applicable			
			0 1 1 0 Factory testing Not applicable								
			0 1 1 Factory testing Not applicable								
			1 0 0 Factory testing Not applicable								
			1 0 0 1 Factory testing Not applicable								
			1 0 1 0 Factory testing Not applicable								
			1	0	1	1	LED mode 3	1536/1536			
			1	1	0	0	LED mode 0	1536/1536			
			1	1	0	1	Factory testing	Not applicable			
			1	1	1	0	LED mode 2	1536 / 1536			
			1	1	1	1	Factory testing	Not applicable			
	FFLOW1#	L	Pulled Flow	d up = no control fo d down =	o force flo eature is	ow contro enabled	e on port 1. If eature on port 1 (and disabled by autory control feature on the control fe				
	FFLOW2#	L	Program force flow control feature on port 2. Pulled up = no force flow control feature on port 2 (default). Flow control feature is enabled and disabled by auto-negotiation. Pulled down = enable force flow control feature on port 2 regardless of auto-negotiation result.								
	LED[1][3]		Program advertise flow control feature for 10/100BaseTX ports during auto-negotiation at reset time. Pulled low = no advertise flow control during auto-negotiation. Pulled high = advertise flow control during auto-negotiation (default).								
	LED[1][2]		Pulled Flow Pulled	d low = 1 control for d high =	no force eature is	flow cont enabled/ orce flow	rol feature on port 3 disabled by auto-ne	ling MII port at reset time. , including MII port (default). gotiation result. ort 3, including MII port regardless			
	LED[1][1]		Rese	rved - us	e float co	onfigurati	on.				

Group	I/O Names	Active Status	Description	on						
	LED[1][0]		option. Pulled low	buffer alloca = 170 buffer n = adaptive	rs (default).	lse the following table to select the				
	LED[2][3]		Programs MAC address aging in the address look-up table at reset time. Aging eliminates old entries from the table. Pulled high = 5 minute aging (default). Pulled low = disable.							
	LED[2][2]			n = enable (d	re enable at reset time. lefault).					
	LED[2][1]			n = enable (d	eack off in half-duplex at res lefault).	et time.				
	LED[2][0]			n = enable (d	e collision drop at reset time lefault).					
	LED[3][3]			al pulldown		and LED[3][3] is being used. If no pull-down is required (floating).				
	LED[3][2:0]		Programs force 100BaseTX mode at reset time. Use the table below to set this mode on the appropriate port. Assuming the corresponding port auto-negotiation is disabled.							
			Signal	Port	Force 10BaseTX	Force 100BaseTX				
			LED[3][2]	Pulled high (default)						
			LED[3][1]	2	Pulled low	Pulled high (default)				
			LED[3][0]	Pulled high (default)						
	MRXD[3:1]					e the table below to set this mode g port auto-negotiation is disabled.				
			Signal	Port	Force Half-Duplex	Force Full-Duplex				
			MRXD[3]	3	Pulled low (default)	Pulled high				
			MRXD[2]	2	Pulled low (default)	Pulled high				
			MRXD[1]	1	Pulled low (default)	Pulled high				
	MRXD0		Pulled high	n = auto-neg	negotiation disable" at reset otiation disable. tiation enable (default).	time.				
	MCOL		Pulled high	n = auto-neg	negotiation disable" at reset otiation disable. tiation enable (default).	time.				
	DISAN3		Programs "port 3 auto-negotiation disable" at reset time. Pulled high = auto-negotiation disable (default) Pulled low = auto-negotiation enable							
	MIIS[1:0]	Н	Selects external MII port operation mode. Use the table below to select the external port mode.							
			MIIS							
			1 0 Selection							
					rnal MII disable (default)					
					everse mode					
					orward mode					
			Н	H 7-wir	e (SNI) mode					

Group	I/O Names	Active Status	Description							
	VMDIS	Н	VLAN Mismatch Discard control. Pulled low = Constrict multicast and broadcast packets to VLAN. Pulled high = Constrict all packets to VLAN (default).							
	PRSV	Н	Pulled low = I	Reserve priority buffers. Pulled low = No buffers reserved (default). Pulled high = Reserve 96 buffers per port for high priority queue.						
	PBASE[2:0]	Н	received that PBASE value equal to the F	has a 802 is compa PBASE va	2.1p tag a ared to th alue, the p	and 802.1 e tag prio packet is	priority tag in 802.1p tag. When a packet is p processing is enabled (Px_1PEN=H), the rity field. If the packet tag is greater than or sent to the higher priority transmit queue ne lower priority queue. (default = 100).			
	PV12 PV13 PV21 PV23 PV31 PV32	Н	VLAN mask between the table below P1_V defined P2_V defined P3_V defined	w to sele as (PV1: as (PV2:	ct VLAN 3, PV12, 3, 1 , PV2	operation 1) 21)	orts are seen from any particular port. Use			
			P[3:1]_V Port	2	1	0	VLAN State			
			1 0 1 1 Ports 1 and 2 in VLAN 1 0 1 Ports 1 and 3 in VLAN 1 1 1 Ports 1, 2 and 3 in VLAN (default)							
			2	0 1 1	1 1 1	1 0 1	Ports 1 and 2 in VLAN Ports 2 and 3 in VLAN Ports 1, 2 and 3 in VLAN (default)			
			3	1 1 1	0 1 1	1 0 1	Ports 1 and 3 in VLAN Ports 2 and 3 in VLAN Ports 1, 2 and 3 in VLAN (default)			
							red for each VLAN. The VLAN configuration All states not listed above are invalid.			
	P[3:1]_1PEN	Н		f the 802. P[3:1]_PI Disable 80	.1p proce P bit. 02.1p pri	essing is do	basis. The enable is from the receive isabled or there is no tag, priority is detedefault).			
	P[3:1]_PP	Н	Selects port r Pulled low = I Pulled high =	_ow priori	ty (defau		e of 802.1p handling.			
	P[3:1]_TAGINS	Н	Inserts 802.1p tag in received packets if not already existent. The priority field is set based on the port P[3:1]_PP bit. For the P[3:1]_PP bit tied low, the priority field is set to 000 and for the P[3:1]_PP bit tied high, the priority field is set to 111. Pulled low = No change to received packet (default). Pulled high = Insert 802.1p tag.							
			Note that if P[3:1]_TAGINS and P[3:1]_TAGRM are both set for the same port, there is no change to the packet.							
	P[3:1]_TAGRM	Н	Removes 802.1p tag in received packets if they exist. Pulled low = No change to received packet (default). Pulled high = Remove 802.1p tag.							
			Note that if P[3:1]_TAGINS and P[3:1]_TAGRM are both set for the same port, there is no change to the packet.							
	P[3:1]_TXQ2	Н	Selects transi queues. Pulled low = 9 Pulled high =	Single tra	nsmit que	eue (defa				

PRSEL[1:0] H Selects queue servicing if using split transmit queues. Use the table below to se discired servicing. Note that this selection effects all split transmit queue ports in same way. PRSEL 1	Group	I/O Names	Active Status	Desc	ription				
1 0 Priority Selection		PRSEL[1:0]	Н	desire	sired servicing. Note that this selection effects all split transmit queue ports in t me way.				
L				l .					
H				L	L	Transr	nit all high priority before low priority (default)		
H				L	Н	Transr	nit high priority at 10:1 ratio		
CTRL X1 Clock External crystal or clock input X2 Clock Used when other polarity of crystal is needed. This is unused for a normal clock RST# L System reset TEST TESTEN H Factory test input - pull low SCANEN H Factory test input - pull low AOUT H Factory test output - leave open MUX[1:2] H Mux[2] Float Tioat Default for factory test purpose 1 Float TX Disable Port 2 Float 1 Power Down Port 3 Special note: all other combinations are not allowed Test[1] Test[1] Test[2] Float Default for factory test purpose Float 1 Final Ploat Default for factory test purpose Float 0 Float 0 Far End Fault Disable Special note: all other combinations are not allowed 2.5V for receiver QND_RX[1:3] Ground for receiver VDD_TX[1:3] Ground for receiver VDD_PLL				Н	L	Transr	nit high priority at 5:1 ratio		
X2				Н	Н	Transr	nit high priority at 2:1 ratio		
RST#	CTRL	X1	Clock	Exter	nal crys	tal or cloc	k input		
TEST TESTEN		X2	Clock	Used	when o	ther polar	ity of crystal is needed. This is unused for a normal clock input.		
SCANEN		RST#	L	Syste	m reset				
AOUT	TEST	TESTEN	Н	Facto	ry test i	nput - pull	low		
MUX[1:2] H Mux[1] Mux[2] Float Float Default for factory test purpose		SCANEN	Н	Facto	ry test i	nput - pull	low		
Float Float Default for factory test purpose 1 Float TX Disable Port 1 0 Float TX Disable Port 2 Float 1 Power Down Port 3 Special note: all other combinations are not allowed TEST[1:2] H Test[1] Test[2] Float Float Default for factory test purpose Float 0 Far End Fault Disable Special note: all other combinations are not allowed. PWR VDD_RX[1:3] Ground for receiver GND_RX[1:3] Ground for receiver VDD_TX[1:3] Ground for transmit circuitry GND_TX[1:3] Ground for transmit circuitry VDD_RCV[1:3] Ground for clock recovery circuitry GND_RCV[1:3] Ground for clock recovery VDD_PLL 2.5V for phase locked loop circuitry GND_PLL Ground for phase locked loop circuitry GND_BG Analog ground VDD_BG 2.5V for core digital circuitry		AOUT	Н	Facto	ry test o	output - lea	ave open		
1 Float TX Disable Port 1 0 Float TX Disable Port 2 Float 1 Power Down Port 3 Special note: all other combinations are not allowed TEST[1:2] H Test[1] Test[2] Float Float Default for factory test purpose Float 0 Far End Fault Disable Special note: all other combinations are not allowed. PWR VDD_RX[1:3] Ground for receiver VDD_TX[1:3] Ground for receiver VDD_TX[1:3] Ground for transmit circuitry GND_RCV[1:3] Ground for transmit circuitry VDD_RCV[1:3] Ground for clock recovery VDD_PLL Ground for phase locked loop circuitry GND_PLL Ground for phase locked loop circuitry GND_ANA Analog ground GND_BG Analog ground VDD_BG 2.5V for core digital circuitry		MUX[1:2]	Н	Mux[1] M	ux[2]			
O Float TX Disable Port 2 Float 1 Power Down Port 3 Special note: all other combinations are not allowed TEST[1:2] H Test[1] Test[2] Float Float Default for factory test purpose Float O Far End Fault Disable Special note: all other combinations are not allowed Special note: all other combinations are not allowed PWR VDD_RX[1:3] Ground for receiver VDD_TX[1:3] Ground for receiver VDD_TX[1:3] Ground for transmit circuitry VDD_RCV[1:3] Ground for transmit circuitry VDD_RCV[1:3] Ground for clock recovery circuitry GND_PLL Ground for clock recovery VDD_PLL Ground for phase locked loop circuitry GND_ANA Analog ground GND_BG Analog ground VDD_BG 2.5V for one digital circuitry VDD_BG 2.5V for core digital circuitry Corea TX Disable Port 2 Power Down Port 3 Power				Float	Flo	oat	Default for factory test purpose		
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Special note: all other combinations are not allowed Test[1] Test[2]				0	Flo	oat	TX Disable Port 2		
TEST[1:2] H Test[1] Test[2] Float Default for factory test purpose Float 0 Far End Fault Disable Special note: all other combinations are not allowed. PWR VDD_RX[1:3] 2.5V for receiver GND_RX[1:3] Ground for receiver VDD_TX[1:3] Ground for transmit circuitry GND_TX[1:3] Ground for transmit circuitry GND_RCV[1:3] Ground for clock recovery circuitry GND_RCV[1:3] Ground for clock recovery VDD_PLL 2.5V for phase locked loop circuitry GND_PLL Ground for phase locked loop circuitry GND_BG Analog ground VDD_BG 2.5V for analog circuitry 2.5V for core digital circuitry				Float	1		Power Down Port 3		
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Special note: all other combinations are not allowed. PWR VDD_RX[1:3] 2.5V for receiver GND_RX[1:3] Ground for receiver VDD_TX[1:3] 2.5V for transmit circuitry GND_TX[1:3] Ground for transmit circuitry VDD_RCV[1:3] 2.5V for clock recovery circuitry GND_RCV[1:3] Ground for clock recovery VDD_PLL 2.5V for phase locked loop circuitry GND_PLL Ground for phase locked loop circuitry GND_ANA Analog ground GND_BG Analog ground VDD_BG 2.5V for analog circuitry 2.5V for core digital circuitry				Float	Flo	oat	Default for factory test purpose		
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GND_TX[1:3] Ground for transmit circuitry VDD_RCV[1:3] 2.5V for clock recovery circuitry GND_RCV[1:3] Ground for clock recovery VDD_PLL 2.5V for phase locked loop circuitry GND_PLL Ground for phase locked loop circuitry GND_ANA Analog ground GND_BG Analog ground VDD_BG 2.5V for analog circuits VDD 2.5V for core digital circuitry		GND_RX[1:3]		Grour	nd for re	ceiver			
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GND_ANA Analog ground GND_BG Analog ground VDD_BG 2.5V for analog circuits VDD 2.5V for core digital circuitry		VDD_PLL		2.5V t	for phas	e locked l	loop circuitry		
GND_BG Analog ground VDD_BG 2.5V for analog circuits VDD 2.5V for core digital circuitry		GND_PLL		Grour	Ground for phase locked loop circuitry				
VDD_BG 2.5V for analog circuits VDD 2.5V for core digital circuitry		GND_ANA		Analo	Analog ground				
VDD 2.5V for core digital circuitry		GND_BG		Analo	nalog ground				
y ,		VDD_BG		2.5V	for anal	og circuits			
VDD_IO 2.5V or 3.3V for MII interface, LEDs and other digital I/O		VDD		2.5V	for core	digital circ	cuitry		
		VDD_IO		2.5V	or 3.3V	for MII inte	erface, LEDs and other digital I/O		
GND Ground for digital circuitry		GND		Grour	nd for di	gital circu	itry		

Pin Configuration



Functional Overview: Physical Layer Transceiver

100BaseTX Transmit

The 100BaseTX transmit function performs parallel to serial conversion, 4B/5B coding, scrambling, NRZ to NRZI conversion, MLT3 encoding and transmission. The circuit starts with a parallel to serial conversion, which converts the data from the MAC into a 125MHz serial bit stream. The data and control stream is then converted into 4B/5B coding followed by a scrambler. The serialized data is further converted from NRZ to NRZI format, then transmitted in MLT3 current output. The output current is set by an external $1\% 3.01 \text{k}\Omega$ resistor for the 1:1 transformer ratio. It has a typical rise/fall time of 4ns and complies to the ANSI TP-PMD standard regarding amplitude balance, overshoot and timing jitters.

100BaseTX Receive

The 100BaseTX receiver function performs adaptive equalization, DC restoration, MLT3 to NRZI conversion, data and clock recovery, NRZI to NRZ conversion, de-scrambling, 4B/5B decoding and serial to parallel conversion. The receiving side starts with the equalization filter to compensate inter-symbol interference (ISI) over the twisted pair cable. Since the amplitude loss and phase distortion is a function of the length of the cable, the equalizer has to adjust its characteristics to optimize the performance. In this design, the variable equalizer will make an initial estimation based on comparisons of incoming signal strength against some known cable characteristics, then tunes itself for optimization. This is an ongoing process and can self adjust against the environmental changes such as temperature variations.

The equalized signal then goes through a DC restoration and data conversion block. The DC restoration circuit is used to compensate for the effect of base line wander and improve the dynamic range. The differential data conversion circuit converts the MLT3 format back to NRZI. The slicing threshold is also adaptive.

The clock recovery circuit extracts the 125MHz clock from the edges of the NRZI signal. This recovered clock is then used to convert the NRZI signal into the NRZ format. The signal is then sent through the de-scrambler followed by the 4B/5B decoder. Finally, the NRZ serial data is provided as the input data to the MAC.

PLL Clock Synthesizer

The KS8993 generates clocks for the external MII and SNI interface based on the interface type selected.

Scrambler/De-scrambler (100BaseTX only)

The purpose of the scrambler is to spread the power spectrum of the signal in order to reduce EMI and baseline wander. The data is scrambled by the use of an 11-bit wide linear feedback shift register (LFSR). This can generate a 2047-bit non-repetitive sequence. The receiver will then de-scramble the incoming data stream with the same sequence at the transmitter.

100BaseFX Operation

100BaseFX operation is very similar to 100BaseTX operation with the differences being that the scrambler/de-scrambler and MLT3 encoder/decoder are bypassed on transmission and reception. In this mode the auto-negotiation feature is bypassed since there is no standard that supports fiber auto-negotiation.

100BaseFX Signal Detection

The physical port runs in 100BaseFX mode if FXSDx >.6V. This signal is referenced to VREFx which is set at 1/2 Vdd but can be overridden by an external level. VREFx can be connected to the "minus" signal of a differential pair coming from the fiber module ("plus connects to FXSDx) used to convey signal detect. When FXSDx is below .6V then 100BaseFX mode is disabled.

100BaseFX Far End Fault

Far end fault occurs when the signal detection is logically false from the receive fiber module. When this occurs, the transmission side signals the other end of the link by sending 84 1's followed by a zero in the idle period between frames.

Far End Fault can be disabled by setting external hardware pin TEST[2]=0 and TEST[1] = float. See "I/O Description" for pin description.

10BaseT Transmit

The output 10BaseT driver is incorporated into the 100BaseT driver to allow transmission with the same magnetic. They are internally wave-shaped and pre-emphasized into outputs with a typical 2.2V amplitude. The harmonic contents are at least 27dB below the fundamental when driven by an all-ones Manchester-encoded signal.

Special note for 10BaseT operation: With an operating voltage of 2.5V, the KS8993 does not always achieve the specified transmit voltage swing greater than or equal to 2.2V as specified by IEEE 802.3. The important factor however is that the KS8993 does adhere to the specified receive signal voltages using the IEEE twisted pair model with a 100Ω load. The transmit voltage swing can be increased to 2.2V or above by increasing the supply voltage to 2.65V if so desired.

10BaseT Receive

On the receive side, input buffer and level detecting squelch circuits are employed. A differential input receiver circuit and a PLL perform the decoding function. The Manchester-encoded data stream is separated into clock signal and NRZ data. A

squelch circuit rejects signals with levels less than 400mV or with short pulse widths in order to prevent noises at the RXP or RXM input from falsely triggering the decoder. When the input exceeds the squelch limit, the PLL locks onto the incoming signal and the KS8993 decodes a data frame. The receiver clock is maintained active during idle periods in between data reception.

Power Management

Power Save Mode

The KS8993 will turn off everything except for the Energy Detect and PLL circuits when the cable is not installed on an individual port basis. In other words, the KS8993 will shutdown most of the internal circuits to save power if there is no link.

An additional features are available:

Transmit Disable for Port 1 set external pin MUX[1] = 1 and MUX[2] = Float

Transmit Disable for Port 2 set external pin MUX[1] = 0 and MUX[2] = Float

Power Down on Port 3 set external pin MUX[1] = float and MUX[2] = 1

See "I/O Description" section for pin description.

LED Mode Selection

Use the following table as a quick reference for setting the LED mode. See MODESEL "I/O Description" section for MODESEL[3:2] usage.

MODESEL[1:0]	LED[1:3]3	LED[1:3]2	LED[1:3]1	LED[1:3]0
00	Link + Act	Collision	FDX	Speed
01	RX Act	TX Act	FDX	Reserved
10	100 / Link / Act	10 / Link / Act	FDX	Collision
11	Link + Act	FDX + Collision	Reserved	Speed

Table 1. LED Mode Selection

Auto-Negotiation

The KS8993 conforms to the auto-negotiation protocol as described by the 802.3 committee. Auto-negotiation allows UTP (Unshielded Twisted Pair) link partners to select the best common mode of operation. In auto-negotiation the link partners advertise capabilities across the link to each other. If auto-negotiation is not supported or the link partner to the KS8993 is forced to bypass auto-negotiation, then the mode is set by observing the signal at the receiver. This is known as parallel mode because while the transmitter is sending auto-negotiation advertisements, the receiver is listening for advertisements or a fixed signal protocol.

The flow for the link set up is depicted below.

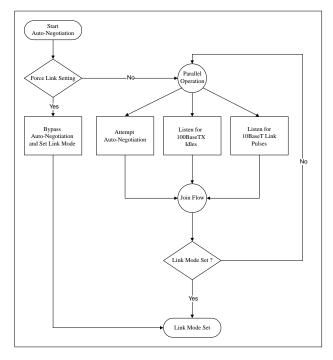


Figure 2. Auto-Negotiation

Functional Overview: Switch Core

Address Look-Up

The internal look-up table stores MAC addresses and their associated information. It contains 1K full CAM with 48-bit address plus switching information. The KS8993 is guaranteed to learn 1K addresses and distinguishes itself from hash-based look-up tables which, depending on the operating environment and probabilities, may not guarantee the absolute number of addresses it can learn.

Learning

The internal look-up engine will update its table with a new entry if the following conditions are met:

- The received packet's SA does not exist in the look-up table.
- The received packet is good; the packet has no receiving errors, and is of legal length.

The look-up engine will insert the qualified SA into the table, along with the port number, time stamp. If the table is full, the last entry of the table will be deleted first to make room for the new entry.

Migration

The internal look-up engine also monitors whether a station is moved. If it happens, it will update the table accordingly. Migration happens when the following conditions are met:

- The received packet's SA is in the table but the associated source port information is different.
- The received packet is good; the packet has no receiving errors, and is of legal length.

The look-up engine will update the existing record in the table with the new source port information.

Aging

The look-up engine will update time stamp information of a record whenever the corresponding SA appears. The time stamp is used in the aging process. If a record is not updated for a period of time, the look-up engine will then remove the record from the table. The look-up engine constantly performs the aging process and will continuously remove aging records. The aging period is approximately $300 \sec 0.5 \pm 75 \sec 0.5$ This feature can be enabled or disabled by external pull-up or pull-down resistors. If aging is disabled and look-up table is full, KS8993 will remove the largest address in the table which has been sorted by the binary search.

Forwarding

The KS8993 will forward packets as follows:

- If the DA look-up results is a "match", the KS8993 will use the destination port information to determine where the packet goes.
- If the DA look-up result is a "miss", the KS8993 will forward the packet to all other ports except the port that received the packet.
- All the multicast and broadcast packets will be forwarded to all other ports except the source port.

The KS8993 will not forward the following packets:

- Error packets. These include framing errors, FCS errors, alignment errors, and illegal size packet errors.
- 802.3x pause frames. The KS8993 will intercept these packets and do the appropriate actions.
- "Local" packets. Based on destination address (DA) look-up. If the destination port from the look-up table matches the port where the packet was from, the packet is defined as "local".

Switching Engine

The KS8993 has a very high performance switching engine to move data to and from the MAC's, packet buffers. It operates in store and forward mode, while the efficient switching mechanism reduces overall latency.

The KS8993 has an internal buffer for frames that is 16kx32 (64kB). This resource is shared between the three ports. Buffer sizing per port can be programmed at system reset time by using the unmanaged program mode (I/O strapping).

Each buffer is sized at 128B and therefore there are a total of 512 buffers available. A per port maximum can be set at 170 buffers (equal allocation). There is also an adaptive mode that reacts to port traffic. In the adaptive mode any given port may use up to 256 buffers provided that the other ports are lightly loaded. In the event of heavier loading on other ports the limit is 170 buffers.

MAC (Media Access Controller) Operation

The KS8993 strictly abides by IEEE 802.3 standard to maximize compatibility and interoperability with other vendors.

Inter Packet Gap (IPG)

If a frame is successfully transmitted, the 96 bit time IPG is measured between the two consecutive MTXEN. If the current packet is experiencing collision, the 96 bit time IPG is measured from MCRS and the next MTXEN.

Back off Algorithm

The KS8993 implements the IEEE Std 802.3 binary exponential back-off algorithm, and optional "aggressive mode" back off. After 16 collisions, the packet will be optionally dropped depending on the chip configuration.

Late Collision

If a transmit packet experiences collisions after 512-bit times of the transmission, the packet will be dropped.

Illegal Frames

The KS8993 will discard illegal size frames defined by the IEEE Std 802.3u, including short frames (less than 64 bytes), long frames (greater than 1522 bytes), and FCS error frames. The KS8993 treats VLAN tagged frames as regular frames and does not perform any VLAN related functions. Switches built with the KS8993's should be treated as a single VLAN domain. KS8993 will drop VLAN frames if the size is larger than 1522 bytes and drop non-VLAN frames if the size is larger than 1518 bytes.

Note that in a special mode, frame lengths of up to 1536 bytes are accepted. This is controlled by MODESEL[3:0]. See "I/O Descriptions" section for more details.

Flow Control

KS8993 supports standard 802.3x flow control frames for full-duplex mode and back-pressure for half-duplex.

Full-Duplex Flow Control (IEEE 802.3x standard)

The flow control capabilities of the KS8993 are enabled based upon the results of the auto-negotiation. During the auto-negotiation, 10/100BaseTX port of KS8993 will advertise this feature to the Link Partner. KS8993 will only establish flow control if the Link Partner has the flow control capability. Since 100BaseFX does not support auto-negotiation, 100BaseFX port will not advertise flow control to the Link Partner.

On the receive side, if the KS8993 receives a pause control frame, the KS8993 will not transmit the next normal frame until the timer, specified in the pause control frame, expires. If another pause frame is received before the current timer expires, the timer will be updated with the new value in the second pause frame. During this period (being flow controlled), only flow control packets from the KS8993 will be transmitted.

On the transmit side, the KS8993 has intelligent and efficient ways to determine when to invoke flow control is based on availability of the system resources, including available buffers, available transmit queues and available receive queues.

The KS8993 will flow control a port, which just received a packet, if the destination port resource is being used up. The KS8993 will issue a flow control frame (XOFF), containing the maximum pause time defined in IEEE standard 802.3x. Once the resource is freed up, the KS8993 will send out the other flow control frame (XON) with zero pause time to turn off the flow control (turn on transmission to the port). A hysterisis feature is provided to prevent flow control mechanism from being activated and deactivated too many times.

The KS8993 will flow control all ports if the receive queue becomes full.

Take a special note that flow control for 100BaseFX or 10/100BaseTX full-duplex can be forced regardless of auto-negotiation result. This force flow control feature on port 1, 2 or 3 can be enabled and disabled via external pin FFLOW#1, FFLOW#2 and LED[1][2] respectively.

Half-Duplex Back Pressure

Half-duplex Back Pressure option (Note: not in 802.3 standards) is also provided. The activation and deactivation conditions are the same as the above in full-duplex mode. If back pressure is required, the KS8993 will send preambles to defer other stations' transmission (carrier sense deference). To avoid jabber and excessive deference defined in 802.3 standard, after a certain time it will discontinue the carrier sense but it will raise the carrier sense quickly. This short silent time (no carrier sense) is to prevent other stations from sending out packets and keeps other stations in carrier sense deferred state. If the port has packets to send during a back pressure situation, the carrier sense type back pressure will be interrupted and those packets will be transmitted instead. If there are no more packets to send, carrier sense type back pressure will be active again until switch resources free up. If a collision occurs, the binary exponential back-off algorithm is skipped and carrier sense is generated immediately, reducing the chance of further colliding and maintaining carrier sense to prevent reception of packets. This scheme is better than collision based back pressure.

VLAN Support

Each port is associated with a 3-bit Port VLAN mask register (PV) (P1_V: (PV13, PV12, 1), P2_V: (PV23, 1, PV21), P3_V: (1, PV32, PV31)). Based on the receiving port's PV, a broadcast packet will be sent to all the ports that have their mask bit set to one, excluding the source port. In other words, broadcast packets will be confined in the VLAN specified in the PV. A unicast packet, which is destined to a port not specified in the PV, could be optionally filtered (depends on the strapped in value during power up, VLAN Mismatch DIScard). The following is a typical set up for a router/switch combo application, in which port 3 is a router port:

P1_V: (1,0,1) P2_V: (1,1,0) P3_V: (1,1,1)

In the above setting, there are two VLANs. VLAN 1 includes ports 1,3 and VLAN 2 includes ports 2, 3. Port 3 belongs to both VLANs. If vmdis = 1, port 1 can never talk to port 2. Port 3 has to route all the traffic across the two VLANs. If vmdis = 0 and there are unicast packets, all ports can talk to all others. If vmdis = 0 and there are multicast packets, those packets are confined in the same VLAN. The router can take advantage of the "vmdis = 0" feature, acting as an agent to handle broadcast/multicast protocol, while leaving unicast switching task to KS8993. For example, port 1 sends an "ARP" for the port 2 MAC address. Since port 2 cannot receive the ARP, the attached router on port 3 will act as an agent and report the MAC address of port 2 to port 1. Then all the unicast traffic between port 1 and port 2 could be switched by KS8993, instead of by the router port. This application could enable "wire speed" switching/routing. This feature is sometimes called "leaky VLAN". This leaky VLAN does improve the system performance by separating broadcast domains. Note KS8993 does not support "duplicated MAC addresses" in different VLANs to save MAC table size.

QoS Priority Support

This feature provides QoS for applications such as VOIP, video conferencing, and mission critical applications. The KS8993 per port transmit queue could be split into two priority queues, high priority and low priority queues. The splitting feature could be optionally per port enabled (using pin Px_TXQ2). If a port is split, high priority packets will be put in the high priority queue. If a port's transmit queue is not split, high priority and low priority packets will be treated equally. There are four priority schemes (selected by pins PRSEL1 and PRSEL0): (1), transmit high priority packets always before low priority packets, i.e. A low priority packet could be transmitted only when the high priority queue is empty. (2), 10/1 ratio, transmit a low priority after every 10 high priority packets transmitted if both queues are busy. (3), 5/1 ratio, (4) 2/1 ratio. Incoming packet priority could be classified in two ways, port-based or 802.1p.

Port based priority: Each port could be individually specified as a high priority receiving port (using pin Px_PP). All the packets received at the high priority receiving port will be marked high priority and sent to the high priority transmit queue if the corresponding queue is split.

802.1p based priority: 802.1p based priority could be enabled by pins Px_1PEN. KS8993 will examine incoming packets to determine whether they are tagged and retrieve the corresponding priority information. The priority field in the VLAN tag is 3 bits wide and is compared against "priority base value specified by pins (PBASE[2:0]). If a received packet has an equal or larger priority value than the "priority base" value, the packet will be put in the high priority transmit queue if the corresponding queue is split. KS8993 can optionally remove or insert priority tagged frame's header (2 bytes of tag protocol identifier 0x8100 and 2 bytes of tag control information). If a transmitting port has its corresponding Px_TAGINS set (meaning tag insertion), the transmitting logic will automatically insert "priority tag" for untagged packets with NULL VLAN ID and its priority value (7 for high priority and 0 for low priority). For already tagged packets, KS8993 will pass the original packet without changing its tag content. If a transmitting port has its corresponding Px_TAGRM set (meaning tag removal), the transmitting logic will automatically remove "802.1q tag". For untagged packets, KS8993 will pass the original packet without changing any content. Either tag insertion or removal will cause CRC recalculation.

MII Interface Operation

The MII (Media Independent Interface) operates in either a forward or reverse mode. In the forward mode, the KS8993 MII acts like a MAC and in the reverse mode, it acts like a PHY device. This interface is specified by the IEEE 802.3 committee and provides a common interface between physical layer and MAC layer devices. There are two distinct groups, one being for transmission and the other for receiving. The table below describes the signals used in this interface in forward and reverse modes.

This interface is a nibble wide data interface and therefore runs at 1/4 the network bit rate (not encoded). Additional signals on the transmit side indicate when data is valid or when an error occurs during transmission. Likewise, the receive side has indicators that convey when the data is valid and without physical layer errors.

For half-duplex operation there is a signal that indicates a collision has occurred during transmission.

Note that the signal MRXER is not provided on the MII interface for the KS8993 for reverse operation and MTXER is not represented for forward mode. Normally this would indicate a receive / transmit error coming from the physical layer /MAC device, but is not appropriate for this configuration. If the connecting device has a MRXER pin, this should be tied low on the other device for reverse or if it has a MTXER pin in the forward mode it should also be tied low on the other device.

Reverse MII Mo	ode Connection		Forward MII Mod	e Connection
External MAC	KS8993 Signal	Description	External PHY	KS8993 Signal
MTXEN	MTXEN	Transmit enable	MTXEN	MRXDV
MTXER	MTXER	Transmit error	MTXER	Not used
MTXD3	MTXD[3]	Transmit data bit 3	MTXD3	MRXD[3]
MTXD2	MTXD[2]	Transmit data bit 2	MTXD2	MRXD[2]
MTXD1	MTXD[1]	Transmit data bit 1	MTXD1	MRXD[1]
MTXD0	MTXD[0]	Transmit data bit 0	MTXD0	MRXD[0]
MTXC	MTXCLK	Transmit clock	MTXC	MTXCLK
MCOL	MCOL	Collision detection	MCOL	MCOLIN
MCRS	MCRS	Carrier sense	MCRS	MCRS
MRXDV	MRXDV	Receive data valid	MRXDV	MTXEN
MRXER	Not used	Receive error	MRXER	MTXER
MRXD3	MRXD[3]	Receive data bit 3	MRXD3	MTXD[3]
MRXD2	MRXD[2]	Receive data bit 2	MRXD2	MTXD[2]
MRXD1	MRXD[1]	Receive data bit 1	MRXD1	MTXD[1]
MRXD0	MRXD[0]	Receive data bit 0	MRXD0	MTXD[0]
MRXC	MRXCLK	Receive clock	MRXC	MRXCLK

Table 2. MII Interconnect

SNI Interface (7-wire) Operation

The SNI (Serial Network Interface) is intended to interface with some controllers used for network layer protocol processing. KS8993 acts like a PHY device to external controllers. This interface can be directly connected to these type of devices. The signals are divided into two groups, one being for transmission and the other being the receive side. The signals involved are described in the table below.

This interface is a bit wide data interface and therefore runs at the network bit rate (not encoded). An additional signal on the transmit side indicates when data is valid. Likewise, the receive side has an indicator that conveys when the data is valid. For half-duplex operation there is a signal that indicates a collision has occurred during transmission.

SNI Signal	Description	KS8993 SNI Signal	KS8993 Input/Output
TXEN	Transmit enable	MTXEN	Input
TXD	Serial transmit data	MTXD[0]	Input
TXC	Transmit clock	MTXCLK	Output
COL	Collision detection	MCOL	Output
CRS	Carrier sense	MRXDV	Output
RXD	Serial receive data	MRXD[0]	Output
RXC	Receive clock	MRXCLK	Output

Table 3. SNI Signal

Absolute Maximum Ratings (Note 1)

Storage Temperature (T_S)–55°C to +150°C

Operating Ratings (Note 2)

+2.35V to +2.75V
+2.35V to +2.75V V or +3.0V to +3.6V
–0°C to +70°C
0 0 10 170 0
42.91°C/W

Electrical Characteristics (Note 4)

 V_{DD} = 2.5V to 2.75V; T_A = 0°C to +70°C; unless noted, **bold** values indicate -40°C $\leq T_A \leq$ +85°C; unless noted.

Symbol	Parameter	Condition	Min	Тур	Max	Units
Total Supp	oly Current (including TX output drive	er current)	'			
I _{DD1}	Normal 100BaseTX			300	330	mA
I _{DD2}	Normal 10BaseT			200	230	mA
TTL Inputs	(V _{DDIO} = 3.3V or 2.5V)	•				
$\overline{V_{IH}}$	Input High Voltage		V _{DD} (I/O) -0.8			V
$\overline{V_{IL}}$	Input Low Voltage				0.8	V
TTL Outpu	ts (V _{DDIO} = 3.3V or 2.5V)	•	•			
V _{OH}	Output High Voltage	I _{OH} = -4mA	V _{DD} (I/O) -0.4			V
V_{OL}	Output Low Voltage	I _{OL} = 4mA			0.4	V
100BaseTX	(Receive	·				
	Error Rate				1.0	1E-8
100BaseTX	C Transmit (measured differentially a	fter 1:1 transformer)	•			
V_{O}	Peak Differential Output Voltage	$50Ω$ from each output to V_{DD}	0.95		1.05	V
V_{IMB}	Output Voltage Imbalance	$50Ω$ from each output to V_{DD}		1.5		%
t _r , t _t	Rise/Fall Time		3		5	ns
100BaseTX	CTransmit (measured differentially a	fter 1:1 transformer)				
V _{SET}	Reference Voltage of ISET			0.75		ns
	Output Jitters	Peak-to-peak		0.7	1.4	ns
10BaseTX	Transmit (measured differentially aft	er 1:1 transformer)				
	Near End Normal Link Pulse	V _{DD} = 2.6V	2.2			V
	Far End Normal Link Pulse	After 100 meters Cat-3 cable	0.5			V
	Far End Output Jitters	After 100 meters Cat-3 cable	-8		8	ns

- Note 1. Exceeding the absolute maximum rating may damage the device.
- Note 2. The device is not guaranteed to function outside its operating rating. Unused inputs must always be tied to an appropriate logic voltage level (Ground to V_{DD}).
- Note 3. No HS (heat spreader) in package.
- Note 4. Specification for packaged product only.

Timing Diagrams

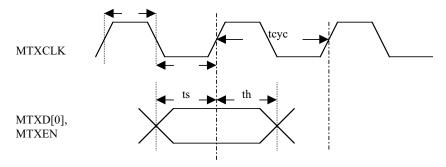


Figure 3. SNI (7-Wire) Input Timing

Symbol	Parameter	Min	Тур	Max	Units
t _{CYC}	Clock Cycle		100		ns
t _S	Set-Up Time	10			ns
t _H	Hold Time	0			ns

Table 4. SNI (7-Wire) Input Timing Parameters

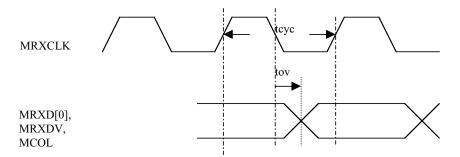


Figure 4. SNI (7-Wire) Output Timing

Symbol	Parameter	Min	Тур	Max	Units
t _{CYC}	Clock Cycle		100		ns
t_{OV}	Output Valid	0	3	6	ns

Table 5. SNI (7-Wire) Output Timing Parameters

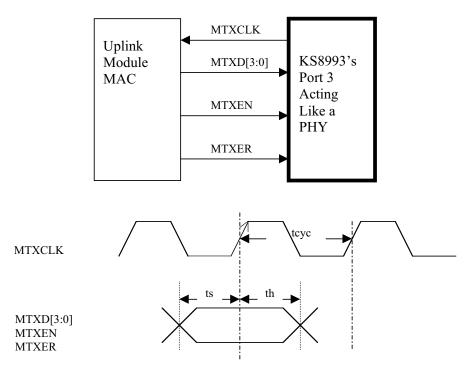


Figure 5. Reverse MII Timing-Receive Data from MII

Symbol	Parameter		Min	Тур	Max	Units
t _{CYC}	Clock Cycle	(100BaseT) (10BaseT)		40 400		ns
t _S	Set-Up Time		10			ns
t _H	Hold Time		0			ns

Table 6. Reverse MII Timing-Receive Data from MII Parameters

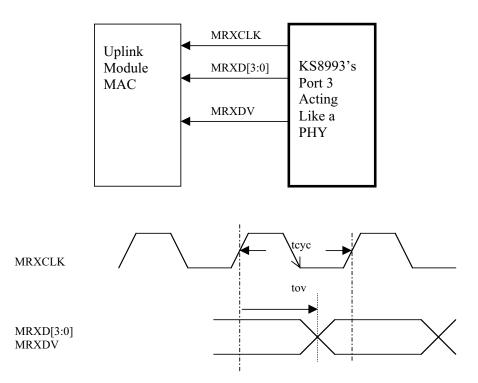


Figure 6. Reverse MII Timing-Transmit Data to MII

Symbol	Parameter		Min	Тур	Max	Units
t _{CYC}	Clock Cycle	(100BaseT) (10BaseT)		40 400		ns
t _{OV}	Output Valid		18	25	28	ns

Table 7. Reverse MII Timing-Transmit Data to MII Parameters

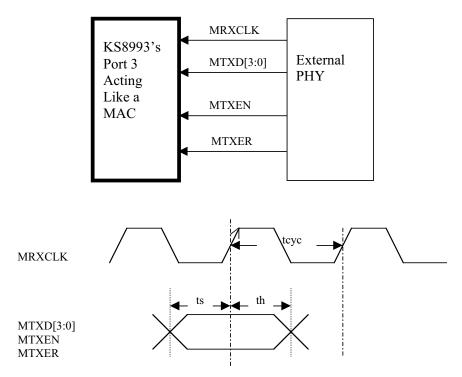


Figure 7. Forward MII Timing-Receive Data from MII

Symbol	Parameter		Min	Тур Мах	Units
t _{CYC}	Clock Cycle	(100BaseT)		40	ns
		(10BaseT)		400	ns
t _S	Set-Up Time		10		ns
t _H	Hold Time		5		ns

Table 8. Forward MII Timing-Receive Data from MII Parameters

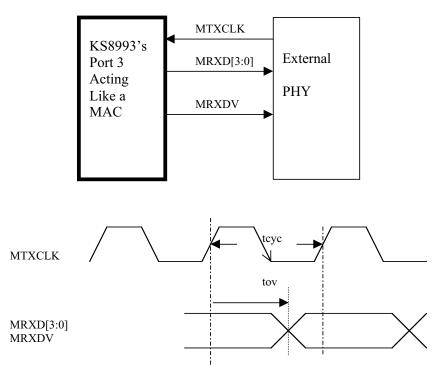


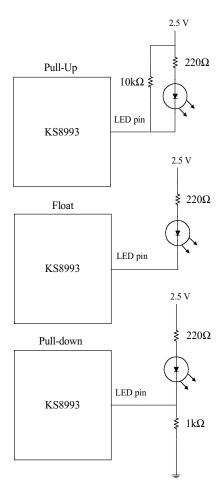
Figure 8. Forward MII Timing-Transmit Data to MII

Symbol	Parameter		Min	Тур	Max	Units
t _{CYC}	Clock Cycle	(100BaseT) (10BaseT)		40 400		ns ns
t _{OV}	Output Valid		7	11	16	ns

Table 9. Forward MII Timing-Transmit Data to MII

Reference Circuit

See "I/O Description" section for pull-up/pull-down and float information.



Reference circuits for unmanaged programming through LED ports

Reset Circuit Diagram

Micrel recommendeds the following discrete reset circuit as shown in Figure 9 when powering up the KS8993 device. For the application where the reset circuit signal comes from another device (e.g., CPU, FPGA, etc), we recommend the reset circuit as shown in Figure 10.

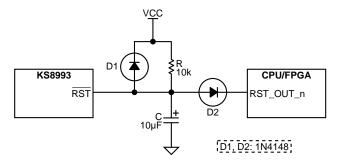


Figure 9. Recommended Reset Circuit.

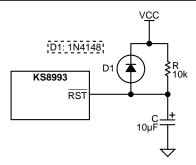


Figure 10. Recommended Circuit for Interfacing with CPU/FPGA Reset

At power-on-reset, R, C, and D1 provide the necessary ramp rise time to reset the Micrel device. The reset out from CPU/FPGA provides warm reset after power up. It is also recommended to power up the VDD core voltage earlier than VDDIO voltage. At worst case, the both VDD core and VDDIO voltages should come up at the same time.

4B/5B Coding

In 100BaseTX and 100BaseFX the data and frame control are encoded in the transmitter (and decoded in the receiver) using a 4B/5B code. The extra code space is required to encode extra control (frame delineation) points. It is also used to reduce run length as well as supply sufficient transitions for clock recovery. The table below provides the translation for the 4B/5B coding.

Code Type	4B Code	5B Code	Value	
Data	0000	11110	Data value 0	
	0001	01001	Data value 1	
	0010	10100	Data value 2	
	0011	10101	Data value 3	
	0100	01010	Data value 4	
	0101	01011	Data value 5	
	0110	01110	Data value 6	
	0111	01111	Data value 7	
	1000	10010	Data value 8	
	1001	10011	Data value 9	
	1010	10110	Data value A	
	1011	10111	Data value B	
	1100	11010	Data value C	
	1101	11011	Data value D	
	1110	11100	Data value E	
	1111	11101	Data value F	
Control	Not defined	11111	Idle	
	0101	11000	Start delimiter part 1	
	0101	10001	Start delimiter part 2	
	Not defined	01101	End delimiter part 1	
	Not defined	00111	End delimiter part 2	
	Not defined	00100	Transmit error	
Invalid	Not defined	00000	Invalid code	
	Not defined	00001	Invalid code	
	Not defined	00010	Invalid code	
	Not defined	00011	Invalid code	
	Not defined	00101	Invalid code	
	Not defined	00110	Invalid code	
	Not defined	01000	Invalid code	
	Not defined	01100	Invalid code	
	Not defined	10000	Invalid code	
	Not defined	11001	Invalid code	

Table 10. 4B/5B Coding

MLT3 Coding

For 100BaseTX operation the NRZI (Non-Return to Zero Invert on ones) signal is line coded as MLT3. The net result of using MLT3 is to reduce the EMI (Electro Magnetic Interference) of the signal over twisted pair media. In NRZI coding, the level changes from high to low or low to high for every "1" bit. For a "0" bit there is no transition. MLT3 line coding transitions through three distinct levels. For every transition of the NRZI signal the MLT3 signal either increments or decrements depending on the current state of the signal. For instance if the MLT3 level is at its lowest point the next two NRZI transitions will change the MLT3 signal initially to the middle level followed by the highest level (second NRZI transition). On the next NRZI change, the MLT3 level will decrease to the middle level. On the following transition of the NRZI signal the MLT3 level will move to the lowest level where the cycle repeats. The diagram below describes the level changes. Note that in the actual 100BaseTX circuit there is a scrambling circuit and that scrambling is not shown in this diagram.

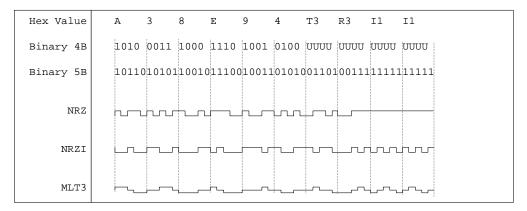


Figure 11. MLT3 coding

The MAC (Media Access Control) fields are described in the table below.

Field	Octect Length	Description	
Preamble/SFD	8	Preamble and Start of Frame Delimiter	
DA	6	48-bit Destination MAC Address	
SA	6	48-bit Source MAC Address	
802.1p tag	4	VLAN and priority tag (optional)	
Length	2	Frame Length	
Protocol/Data	46 to 1500	Higher Layer Protocol and Frame Data	
Frame CRC	4	32-bit Cyclical Redundancy Check	
ESD	1	End of Stream Delimiter	
Idle	Variable	Inter Frame Idles	

Table 11. MAC Frame for 802.3

802.1q VLAN and 802.1p Priority Frame

The 3-bit of 802.1p priority is embedded into the 802.1q VLAN frame as described below:

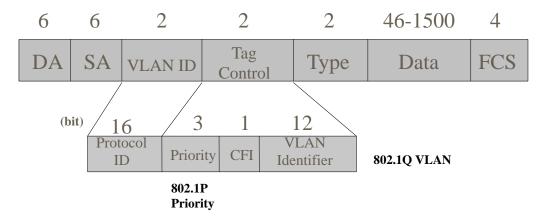


Figure 12. 802.1p and 802.1q Frame Format

Selection of Isolation Transformer^(Note 1)

One simple 1:1 isolation transformer is needed at the line interface. An isolation transformer with integrated common-mode choke is recommended for exceeding FCC requirements. The following table gives recommended transformer characteristics.

Characteristics Name	Value	Test Condition	
Turns Ratio	1 CT : 1 CT		
Open-Circuit Inductance (min.)	350μΗ	100mV, 100 KHz, 8mA	
Leakage Inductance (max.)	0.4μΗ	1MHz (min.)	
Inter-Winding Capacitance (max.)	12pF		
D.C. Resistance (max.)	0.9Ω		
Insertion Loss (max.)	1.0dB	0MHz to 65MHz	
HIPOT (min.)	1500Vrms		

Note 1. The IEEE 802.3u standard for 100BaseTX assumes a transformer loss of 0.5 dB. For the transmit line transformer, insertion loss of up to 1.3dB can be compensated by increasing the line drive current by means of reducing the ISET resistor value.

Selection of Reference Crystal

An oscillator or crystal with the following typical characteristics is recommended.

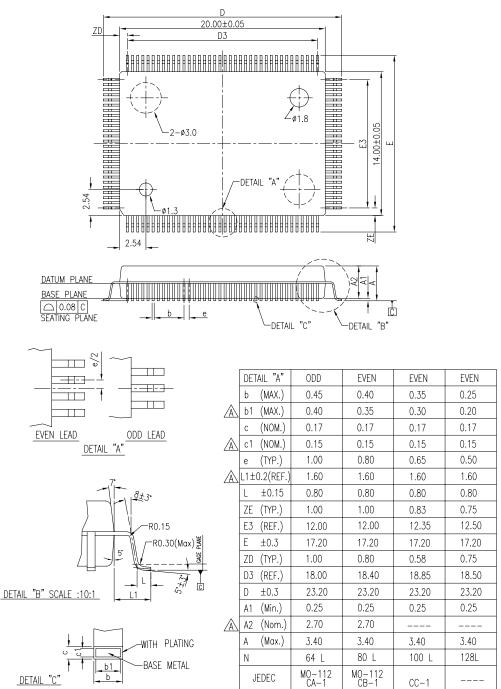
Characteristics Name	Value	Units
Frequency	25.00000	MHz
Frequency Tolerance (max.)	±100	ppm

The following transformer vendors provide pin-to-pin compatible parts for Micrel's device:

Туре	Quad		Single	
	Vendor	Part	Vendor	Part
Transformer only	Pulse	H1060	Pulse	H1012
	YCL	PH406080	YCL	20PMT04
	Trans-Power	HB826-10	Trans-Power	HB614-1-LP
Integrated RJ45 and Transformer	Trans-Power	RJG4-754-C-NL	Trans-Power	RJ754-C-NL

Table 12. Qualified Transformer Lists

Package Information



128-Pin PQFP (PQ)

MICREL, INC. 2180 FORTUNE DRIVE SAN JOSE, CA 95131 USA

TEL + 1 (408) 944-0800 FAX + 1 (408) 474-1000 WEB http://www.micrel.com

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