

#### **Table 1. Maximum Ratings**

Rating	Symbol	Value	Unit
Drain-Source Voltage	V <sub>DSS</sub>	-0.5, +40	Vdc
Gate-Source Voltage	V <sub>GS</sub>	±20	Vdc
Drain Current — Continuous	I <sub>D</sub>	4	Adc
Total Device Dissipation @ T <sub>C</sub> = 25°C <sup>(1)</sup> Derate above 25°C	P <sub>D</sub>	62.5 0.5	W W/°C
Storage Temperature Range	T <sub>stg</sub>	- 65 to +150	°C
Operating Junction Temperature	TJ	150	°C

#### **Table 2. Thermal Characteristics**

Characteristic	Symbol	Value <sup>(2)</sup>	Unit	
Thermal Resistance, Junction to Case	$R_{ extsf{ heta}JC}$	2	°C/W	

#### Table 3. Moisture Sensitivity Level

Test Methodology		Rating Package Peak Temperature	
Per JESD22-A113, IPC/JEDEC J-STD-020		260	°C

1. Calculated based on the formula  $P_D = \frac{T_J - T_C}{R_0 + C}$ 

2. MTTF calculator available at <a href="http://www.freescale.com/rf">http://www.freescale.com/rf</a>. Select Software & Tools/Development Tools/Calculators to access MTTF calculators by product.

NOTE - **CAUTION** - MOS devices are susceptible to damage from electrostatic charge. Reasonable precautions in handling and packaging MOS devices should be observed.

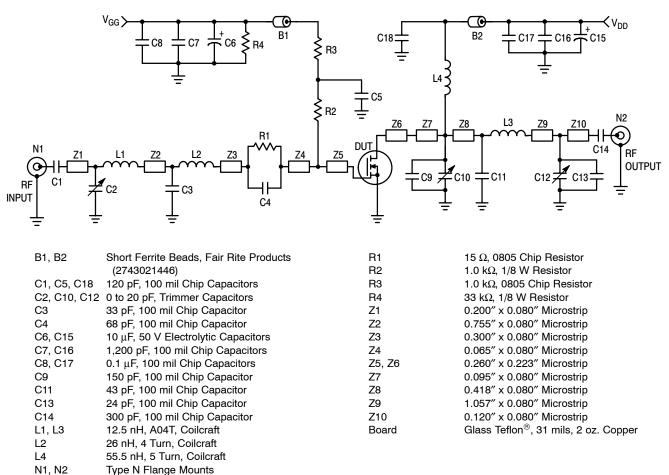




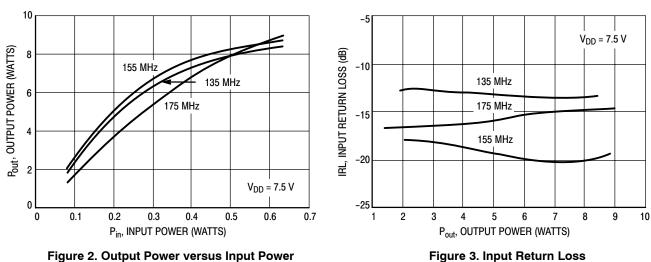
Characteristic	Symbol	Min	Тур	Max	Unit
Off Characteristics					
Zero Gate Voltage Drain Current (V <sub>DS</sub> = 35 Vdc, V <sub>GS</sub> = 0)	I <sub>DSS</sub>		_	1	μAdc
Gate-Source Leakage Current (V <sub>GS</sub> = 10 Vdc, V <sub>DS</sub> = 0)	I <sub>GSS</sub>	_	_	1	μAdc
On Characteristics			1		
Gate Threshold Voltage (V <sub>DS</sub> = 7.5 Vdc, I <sub>D</sub> = 170 μA)	V <sub>GS(th)</sub>	1	1.6	2.1	Vdc
Drain-Source On-Voltage (V <sub>GS</sub> = 10 Vdc, I <sub>D</sub> = 1 Adc)	V <sub>DS(on)</sub>	_	0.4	_	Vdc
Dynamic Characteristics			1		
Input Capacitance (V <sub>DS</sub> = 7.5 Vdc, V <sub>GS</sub> = 0, f = 1 MHz)	C <sub>iss</sub>		100		pF
Output Capacitance $(V_{DS} = 7.5 \text{ Vdc}, V_{GS} = 0, f = 1 \text{ MHz})$	C <sub>oss</sub>		53	_	pF
Reverse Transfer Capacitance $(V_{DS} = 7.5 \text{ Vdc}, V_{GS} = 0, f = 1 \text{ MHz})$	C <sub>rss</sub>	_	8	_	pF
Functional Tests (In Freescale Test Fixture)			1		
Common-Source Amplifier Power Gain $(V_{DD} = 7.5 \text{ Vdc}, P_{out} = 8 \text{ Watts}, I_{DQ} = 150 \text{ mA}, f = 175 \text{ MHz})$	G <sub>ps</sub>		13	_	dB
Drain Efficiency (V <sub>DD</sub> = 7.5 Vdc, P <sub>out</sub> = 8 Watts, I <sub>DQ</sub> = 150 mA, f = 175 MHz)	η		70	_	%

# MRF1511NT1

	$\mathbf{A}$	





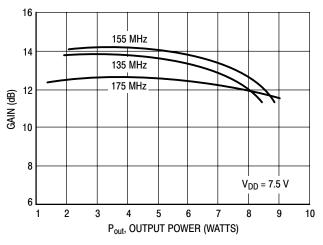


# **TYPICAL CHARACTERISTICS, 135 - 175 MHz**

versus Output Power



**TYPICAL CHARACTERISTICS, 135 - 175 MHz** 





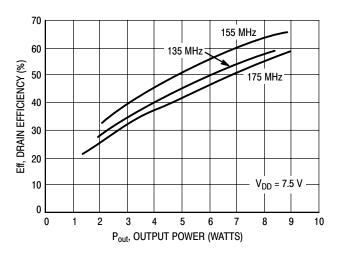


Figure 5. Drain Efficiency versus Output Power

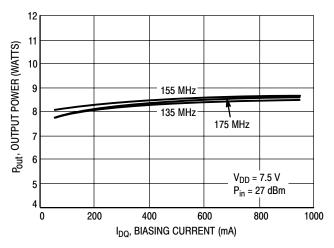


Figure 6. Output Power versus Biasing Current

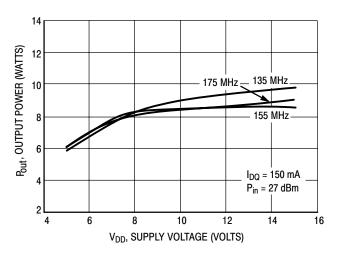


Figure 8. Output Power versus Supply Voltage

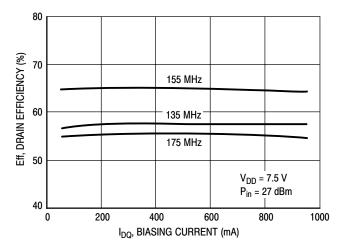


Figure 7. Drain Efficiency versus Biasing Current

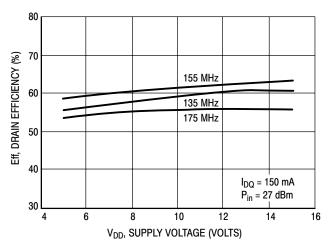
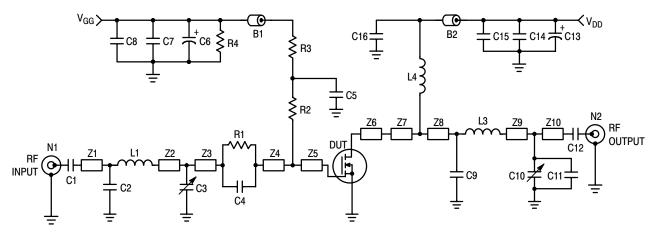


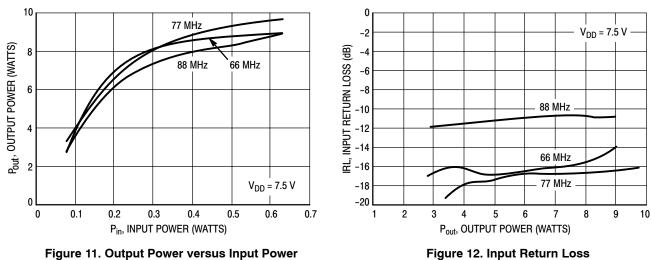
Figure 9. Drain Efficiency versus Supply Voltage

## MRF1511NT1



B1, B2	Short Ferrite Beads, Fair Rite Products	N1, N2	Type N Flange Mounts
	(2743021446)	R1	15 Ω, 0805 Chip Resistor
C1, C12	330 pF, 100 mil Chip Capacitors	R2	51 Ω, 1/2 W Resistor
C2	43 pF, 100 mil Chip Capacitor	R3	100 Ω, 0805 Chip Resistor
C3, C10	0 to 20 pF, Trimmer Capacitors	R4	33 kΩ, 1/8 W Resistor
C4	24 pF, 100 mil Chip Capacitor	Z1	0.136" x 0.080" Microstrip
C5, C16	120 pF, 100 mil Chip Capacitors	Z2	0.242" x 0.080" Microstrip
C6, C13	10 µF, 50 V Electrolytic Capacitors	Z3	1.032" x 0.080" Microstrip
C7, C14	1,200 pF, 100 mil Chip Capacitors	Z4	0.145" x 0.080" Microstrip
C8, C15	0.1 µF, 100 mil Chip Capacitors	Z5, Z6	0.260" x 0.223" Microstrip
C9	380 pF, 100 mil Chip Capacitor	Z7	0.134" x 0.080" Microstrip
C11	75 pF, 100 mil Chip Capacitor	Z8	0.490" x 0.080" Microstrip
L1	82 nH, Coilcraft	Z9	0.872" x 0.080" Microstrip
L2	55.5 nH, 5 Turn, Coilcraft	Z10	0.206" x 0.080" Microstrip
L3	39 nH, 6 Turn, Coilcraft	Board	Glass Teflon <sup>®</sup> , 31 mils, 2 oz. Copper

Figure 10. 66 - 88 MHz Broadband Test Circuit

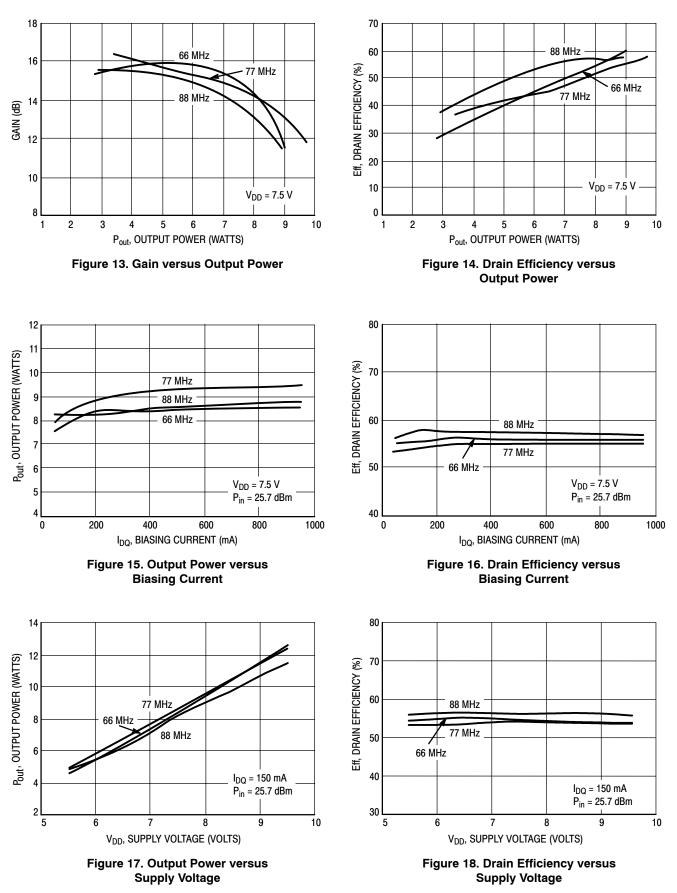


# **TYPICAL CHARACTERISTICS, 66 - 88 MHz**

Figure 12. Input Return Loss versus Output Power



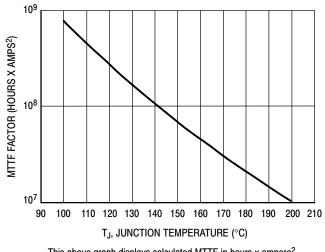
**TYPICAL CHARACTERISTICS, 66 - 88 MHz** 



MRF1511NT1

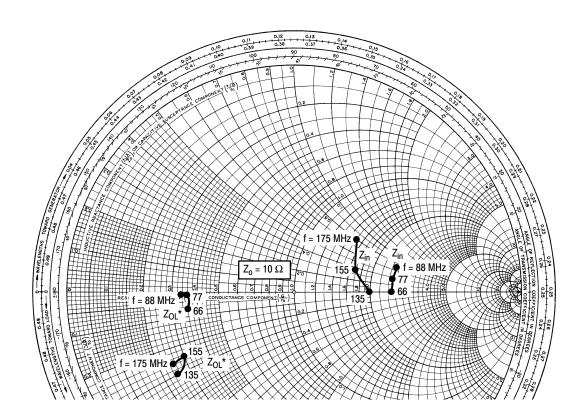


# **TYPICAL CHARACTERISTICS**



This above graph displays calculated MTTF in hours x ampere<sup>2</sup> drain current. Life tests at elevated temperatures have correlated to better than ±10% of the theoretical prediction for metal failure. Divide MTTF factor by  $I_D{}^2$  for MTTF in a particular application.





 $V_{DD}$  = 7.5 V,  $I_{DQ}$  = 150 mA,  $P_{out}$  = 8 W

f MHz	Z <sub>in</sub> Ω	<b>Ζ<sub>ΟL</sub>*</b> Ω
135	20.1 -j0.5	2.53 -j2.61
155	17.0 +j3.6	3.01 -j2.48
175	15.2 +j7.9	2.52 -j3.02

- $Z_{in}$  = Complex conjugate of source impedance with parallel 15  $\Omega$ resistor and 68 pF capacitor in series with gate. (See Figure 1).

 $V_{DD}$  = 7.5 V,  $I_{DQ}$  = 150 mA,  $P_{out}$  = 8 W

f MHz	Z <sub>in</sub> Ω	<b>Ζ<sub>ΟL</sub>*</b> Ω
66	25.3 -j0.31	3.62 -j0.751
77	25.6 +j3.62	3.59 -j0.129
88	26.7 +j6.79	3.37 -j0.173

 $Z_{in} = Complex conjugate of source$  $impedance with parallel 15 \Omega$ resistor and 24 pF capacitor inseries with gate. (See Figure 10).

Note:  $Z_{OL}^*$  was chosen based on tradeoffs between gain, drain efficiency, and device stability.

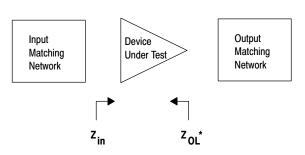


Figure 20. Series Equivalent Input and Output Impedance



	I <sub>DQ</sub> = 150 mA							
f	S	11	S	S <sub>21</sub>		12	S <sub>22</sub>	
MHz	S <sub>11</sub>	$\angle \phi$	S <sub>21</sub>	$\angle \phi$	S <sub>12</sub>	$\angle \phi$	S <sub>22</sub>	$\angle \phi$
30	0.88	-165	18.92	95	0.015	8	0.84	- 169
50	0.88	-171	11.47	91	0.016	-5	0.84	- 173
100	0.87	-175	5.66	85	0.016	-7	0.84	- 176
150	0.87	-176	3.75	82	0.015	-5	0.85	- 176
200	0.87	-177	2.78	78	0.014	-6	0.84	- 176
250	0.87	-177	2.16	75	0.014	-10	0.85	- 176
300	0.88	-177	1.77	72	0.012	-17	0.86	- 176
350	0.88	-177	1.49	69	0.013	-11	0.86	- 176
400	0.88	-177	1.26	66	0.013	-17	0.87	- 175
450	0.88	-177	1.08	64	0.011	-20	0.87	- 175
500	0.89	-176	0.96	63	0.012	-20	0.88	- 175

# Table 5. Common Source Scattering Parameters ( $V_{DD}$ = 7.5 Vdc)

I<sub>DQ</sub> = 800 mA

f	S	11	S	21	S	12	S	22
MHz	S <sub>11</sub>	$\angle \phi$	S <sub>21</sub>	$\angle \phi$	S <sub>12</sub>	$\angle \phi$	S <sub>22</sub>	$\angle \phi$
30	0.89	-166	18.89	95	0.014	10	0.85	-170
50	0.88	-172	11.44	91	0.015	8	0.84	-174
100	0.87	-175	5.65	86	0.016	-2	0.85	-176
150	0.87	-177	3.74	82	0.014	-8	0.84	-177
200	0.87	-177	2.78	78	0.013	-18	0.85	-177
250	0.88	-177	2.16	75	0.012	-11	0.85	-176
300	0.88	-177	1.77	73	0.015	-15	0.86	-176
350	0.88	-177	1.50	70	0.009	-7	0.87	-176
400	0.88	-177	1.26	67	0.012	-3	0.87	-176
450	0.88	-177	1.09	65	0.012	-18	0.87	- 175
500	0.89	-177	0.97	64	0.009	-10	0.88	-175

## I<sub>DQ</sub> = 1.5 A

f	S	11	S	21	S <sub>12</sub>		S <sub>22</sub>	
MHz	S <sub>11</sub>	$\angle \phi$	S <sub>21</sub>	$\angle \phi$	S <sub>12</sub>	$\angle \phi$	S <sub>22</sub>	$\angle \phi$
30	0.90	-168	17.89	95	0.013	2	0.86	- 172
50	0.89	-173	10.76	91	0.013	3	0.86	- 175
100	0.88	-176	5.32	86	0.014	- 19	0.86	- 177
150	0.88	-177	3.53	83	0.013	-6	0.86	- 177
200	0.88	-177	2.63	80	0.011	-4	0.86	- 177
250	0.88	-178	2.05	77	0.012	-14	0.86	- 177
300	0.88	-177	1.69	75	0.013	-2	0.87	- 177
350	0.89	-177	1.43	72	0.010	-9	0.87	- 176
400	0.89	-177	1.22	70	0.014	-3	0.88	- 176
450	0.89	-177	1.06	68	0.011	-8	0.88	- 176
500	0.89	-177	0.94	67	0.011	-15	0.88	- 176

RF Device Data

## **DESIGN CONSIDERATIONS**

This device is a common-source, RF power, N-Channel enhancement mode, Lateral <u>Metal-Oxide Semiconductor</u> <u>Field-Effect Transistor (MOSFET)</u>. Freescale Application Note AN211A, "FETs in Theory and Practice", is suggested reading for those not familiar with the construction and characteristics of FETs.

This surface mount packaged device was designed primarily for VHF and UHF portable power amplifier applications. Manufacturability is improved by utilizing the tape and reel capability for fully automated pick and placement of parts. However, care should be taken in the design process to insure proper heat sinking of the device.

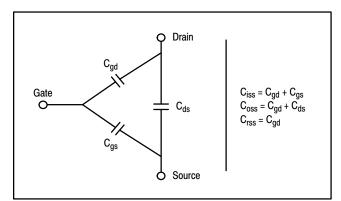
The major advantages of Lateral RF power MOSFETs include high gain, simple bias systems, relative immunity from thermal runaway, and the ability to withstand severely mismatched loads without suffering damage.

#### **MOSFET CAPACITANCES**

The physical structure of a MOSFET results in capacitors between all three terminals. The metal oxide gate structure determines the capacitors from gate-to-drain ( $C_{gd}$ ), and gate-to-source ( $C_{gs}$ ). The PN junction formed during fabrication of the RF MOSFET results in a junction capacitance from drain-to-source ( $C_{ds}$ ). These capacitances are characterized as input ( $C_{iss}$ ), output ( $C_{oss}$ ) and reverse transfer ( $C_{rss}$ ) capacitances on data sheets. The relationships between the inter-terminal capacitances and those given on data sheets are shown below. The  $C_{iss}$  can be specified in two ways:

- 1. Drain shorted to source and positive voltage at the gate.
- 2. Positive voltage of the drain in respect to source and zero volts at the gate.

In the latter case, the numbers are lower. However, neither method represents the actual operating conditions in RF applications.



### **DRAIN CHARACTERISTICS**

One critical figure of merit for a FET is its static resistance in the full-on condition. This on-resistance,  $R_{DS(on)}$ , occurs in the linear region of the output characteristic and is specified at a specific gate-source voltage and drain current. The drain-source voltage under these conditions is termed  $V_{DS(on)}$ . For MOSFETs,  $V_{DS(on)}$  has a positive temperature coefficient at high temperatures because it contributes to the power dissipation within the device.

 $\mathsf{BV}_{\mathsf{DSS}}$  values for this device are higher than normally required for typical applications. Measurement of  $\mathsf{BV}_{\mathsf{DSS}}$  is not recommended and may result in possible damage to the device.

#### **GATE CHARACTERISTICS**

The gate of the RF MOSFET is a polysilicon material, and is electrically isolated from the source by a layer of oxide. The DC input resistance is very high - on the order of  $10^9 \Omega$ — resulting in a leakage current of a few nanoamperes.

Gate control is achieved by applying a positive voltage to the gate greater than the gate-to-source threshold voltage,  $V_{GS(th)}$ .

**Gate Voltage Rating** — Never exceed the gate voltage rating. Exceeding the rated  $V_{GS}$  can result in permanent damage to the oxide layer in the gate region.

**Gate Termination** — The gates of these devices are essentially capacitors. Circuits that leave the gate open-circuited or floating should be avoided. These conditions can result in turn-on of the devices due to voltage build-up on the input capacitor due to leakage currents or pickup.

**Gate Protection** — These devices do not have an internal monolithic zener diode from gate-to-source. If gate protection is required, an external zener diode is recommended. Using a resistor to keep the gate-to-source impedance low also helps dampen transients and serves another important function. Voltage transients on the drain can be coupled to the gate through the parasitic gate-drain capacitance. If the gate-to-source impedance and the rate of voltage change on the drain are both high, then the signal coupled to the gate may be large enough to exceed the gate-threshold voltage and turn the device on.

#### DC BIAS

Since this device is an enhancement mode FET, drain current flows only when the gate is at a higher potential than the source. RF power FETs operate optimally with a quiescent drain current ( $I_{DQ}$ ), whose value is application dependent. This device was characterized at  $I_{DQ} = 150$  mA, which is the suggested value of bias current for typical applications. For special applications such as linear amplification,  $I_{DQ}$  may have to be selected to optimize the critical parameters.

The gate is a dc open circuit and draws no current. Therefore, the gate bias circuit may generally be just a simple resistive divider network. Some special applications may require a more elaborate bias system.

#### GAIN CONTROL

Power output of this device may be controlled to some degree with a low power dc control signal applied to the gate, thus facilitating applications such as manual gain control, ALC/AGC and modulation systems. This characteristic is very dependent on frequency and load line.



## MOUNTING

The specified maximum thermal resistance of  $2^{\circ}$ C/W assumes a majority of the 0.065" x 0.180" source contact on the back side of the package is in good contact with an appropriate heat sink. As with all RF power devices, the goal of the thermal design should be to minimize the temperature at the back side of the package. Refer to Freescale Application Note AN4005/D, "Thermal Management and Mounting Method for the PLD-1.5 RF Power Surface Mount Package," and Engineering Bulletin EB209/D, "Mounting Method for RF Power Leadless Surface Mount Transistor" for additional information.

## AMPLIFIER DESIGN

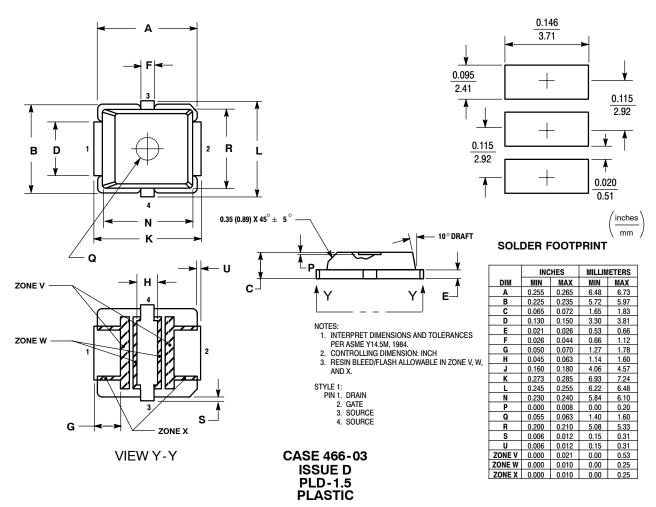
Impedance matching networks similar to those used with bipolar transistors are suitable for this device. For examples see Freescale Application Note AN721, "Impedance Matching Networks Applied to RF Power Transistors." Large-signal impedances are provided, and will yield a good first pass approximation.

Since RF power MOSFETs are triode devices, they are not unilateral. This coupled with the very high gain of this device yields a device capable of self oscillation. Stability may be achieved by techniques such as drain loading, input shunt resistive loading, or output to input feedback. The RF test fixture implements a parallel resistor and capacitor in series with the gate, and has a load line selected for a higher efficiency, lower gain, and more stable operating region.

Two-port stability analysis with this device's S-parameters provides a useful tool for selection of loading or feedback circuitry to assure stable operation. See Free-scale Application Note AN215A, "RF Small-Signal Design Using Two-Port Parameters" for a discussion of two port network theory and stability.



# PACKAGE DIMENSIONS





## PRODUCT DOCUMENTATION, TOOLS AND SOFTWARE

Refer to the following documents to aid your design process.

## **Application Notes**

- AN211A: Field Effect Transistors in Theory and Practice
- AN215A: RF Small-Signal Design Using Two-Port Parameters
- AN721: Impedance Matching Networks Applied to RF Power Transistors
- AN4005: Thermal Management and Mounting Method for the PLD 1.5 RF Power Surface Mount Package

## **Engineering Bulletins**

• EB212: Using Data Sheet Impedances for RF LDMOS Devices

## Software

• Electromigration MTTF Calculator

For Software and Tools, do a Part Number search at http://www.freescale.com, and select the "Part Number" link. Go to the Software & Tools tab on the part's Product Summary page to download the respective tool.

## **REVISION HISTORY**

The following table summarizes revisions to this document.

Revision	Date	Description
7	June 2008	<ul> <li>Corrected specified performance values for power gain and efficiency on p. 1 to match typical performance values in the functional test table on p. 2</li> <li>Added Product Documentation and Revision History, p. 13</li> </ul>
8	June 2009	<ul> <li>Modified data sheet to reflect MSL rating change from 1 to 3 as a result of the standardization of packing process as described in Product and Process Change Notification number, PCN13516, p. 1</li> <li>Added Electromigration MTTF Calculator availability to Product Documentation, Tools and Software, p. 13</li> </ul>

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